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LM2677 SIMPLE SWITCHER® High Efficiency 5A Step-Down Voltage Regulator with Sync

Check for Samples: LM2677

FEATURES

- Efficiency up to 92%
- Simple and easy to design with (using off-theshelf external components)
- 100 mΩ DMOS output switch
- 3.3V, 5V and 12V fixed output and adjustable (1.2V to 37V) versions
- 50µA standby current when switched OFF
- ±2%maximum output tolerance over full line and load conditions
- Wide input voltage range: 8V to 40V
- External Sync clock capability (280KHz to 400KHz)
- 260 KHz fixed frequency internal oscillator

-40 to +125°C operating junction temperature range

APPLICATIONS

- Simple to design, high efficiency (>90%) stepdown switching regulators
- Efficient system pre-regulator for linear voltage regulators
- · Battery chargers
- Communications and radio equipment regulator with synchronized clock frequency

DESCRIPTION

The LM2677 series of regulators are monolithic integrated circuits which provide all of the active functions for a step-down (buck) switching regulator capable of driving up to 5A loads with excellent line and load regulation characteristics. High efficiency (>90%) is obtained through the use of a low ON-resistance DMOS power switch. The series consists of fixed output voltages of 3.3V, 5V and 12V and an adjustable output version.

The SIMPLE SWITCHER™ concept provides for a complete design using a minimum number of external components. The switching clock frequency can be provided by an internal fixed frequency oscillator (260KHz) or from an externally provided clock in the range of 280KHz to 400Khz which allows the use of physically smaller sized components. A family of standard inductors for use with the LM2677 are available from several manufacturers to greatly simplify the design process. The external Sync clock provides direct and precise control of the output ripple frequency for consistent filtering or frequency spectrum positioning.

The LM2677 series also has built in thermal shutdown, current limiting and an ON/OFF control input that can power down the regulator to a low 50μ A quiescent current standby condition. The output voltage is guaranteed to a $\pm 2\%$ tolerance.

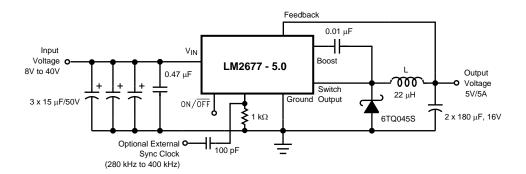
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Typical Application



Connection Diagrams and Ordering Information

Top View 7 on/off 6 FEEDBACK 15 SYNC 13 GROUND 13 C BOOST 12 INPUT 0 1 SWITCH OUTPUT

Figure 1. TO-263 Package

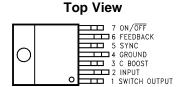


Figure 2. TO-220 Package

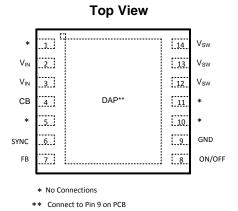


Figure 3. LLP-14 Package





These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings (1)

45V
-0.1V to 6V
-1V to V _{IN}
V _{SW} + 8V
-0.3V to 14V
Internally Limited
2 kV
−65°C to 150°C
4 sec, 260°C
10 sec, 240°C
75 sec, 219°C
· · · · · · · · · · · · · · · · · · ·

- (1) Absolute Maximum Ratings are limits beyond which damage to the device may occur. Operating Ratings indicate conditions under which of the device is guaranteed. Operating Ratings do not imply guaranteed performance limits. For guaranteed performance limits and associated test condition, see the electrical Characteristics tables.
- (2) The absolute maximum specification of the 'Switch Voltage to Ground' applies to DC voltage. An extended negative voltage limit of -10V applies to a pulse of up to 20 ns, -6V of 60 ns and -3V of up to 100 ns.
- (3) ESD was applied using the human-body model, a 100pF capacitor discharged through a 1.5 kΩ resistor into each pin.

Operating Ratings

Supply Voltage	8V to 40V
Junction Temperature Range (T _J)	−40°C to 125°C



LM2677-3.3 Electrical Characteristics

Limits appearing in **bold type face** apply over the entire junction temperature range of operation, -40° C to 125°C. Specifications appearing in normal type apply for $T_A = T_J = 25^{\circ}$ C. Sync pin open circuited.

Symbol	Parameter	Conditions	Typ ⁽¹⁾	Min ⁽²⁾	Max ⁽²⁾	Units
V _{OUT}	Output Voltage	$V_{IN} = 8V \text{ to } 40V, 100\text{mA} \le I_{OUT} \le 5A$	3.3	3.234 3.201	3.366 3.399	٧
η	Efficiency	$V_{IN} = 12V$, $I_{LOAD} = 5A$	82			%

(1) Typical values are determined with $T_A = T_J = 25$ °C and represent the most likely norm.

(2) All limits are guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% tested during production with T_A = T_J = 25°C. All limits at temperature extremes are guaranteed via correlation using standard standard Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

LM2677-5.0 Electrical Characteristics

Symbol	Parameter	Conditions	Typ ⁽¹⁾	Min ⁽²⁾	Max ⁽²⁾	Units
V _{OUT}	Output Voltage	$V_{IN} = 8V \text{ to } 40V, 100\text{mA} \le I_{OUT} \le 5A$	5.0	4.900 4.850	5.100 5.150	V
η	Efficiency	$V_{IN} = 12V$, $I_{LOAD} = 5A$	84			%

(1) Typical values are determined with $T_A = T_J = 25$ °C and represent the most likely norm.

(2) All limits are guaranteed at room temperature (standard type face) and at **temperature extremes (bold type face)**. All room temperature limits are 100% tested during production with T_A = T_J = 25°C. All limits at temperature extremes are guaranteed via correlation using standard standard Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

LM2677-12 Electrical Characteristics

Symbol	Parameter	Conditions	Typ ⁽¹⁾	Min ⁽²⁾	Max ⁽²⁾	Units
V _{OUT}	Output Voltage	$V_{IN} = 15V \text{ to } 40V, 100\text{mA} \le I_{OUT} \le 5A$	12	11.76 11.64	12.24 12.36	\
η	Efficiency	$V_{IN} = 24V$, $I_{LOAD} = 5A$	92			%

(1) Typical values are determined with $T_A = T_J = 25^{\circ}\text{C}$ and represent the most likely norm.

(2) All limits are guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% tested during production with T_A = T_J = 25°C. All limits at temperature extremes are guaranteed via correlation using standard standard Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

LM2677-ADJ Electrical Characteristics

Symbol	Parameter	Conditions	Typ ⁽¹⁾	Min ⁽²⁾	Max ⁽²⁾	Units
V_{FB}	Feedback Voltage	V_{IN} = 8V to 40V, 100mA \leq I_{OUT} \leq 5A V_{OUT} Programmed for 5V	1.21	1.186 1.174	1.234 1.246	V
η	Efficiency	$V_{IN} = 12V$, $I_{LOAD} = 5A$	84			%

(1) Typical values are determined with $T_A = T_J = 25^{\circ}C$ and represent the most likely norm.

(2) All limits are guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% tested during production with T_A = T_J = 25°C. All limits at temperature extremes are guaranteed via correlation using standard standard Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).



All Output Voltage Versions Electrical Characteristics

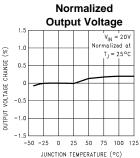
Limits appearing in **bold type face** apply over the entire junction temperature range of operation, -40° C to 125°C. Specifications appearing in normal type apply for $T_A = T_J = 25^{\circ}$ C. Unless otherwise specified $V_{IN}=12V$ for the 3.3V, 5V and Adjustable versions and $V_{IN}=24V$ for the 12V version, Sync pin open circuited..

Symbol	Parameter	Conditions	Тур	Min	Max	Units
DEVICE F	PARAMETERS					
IQ	Quiescent Current	V _{FEEDBACK} = 8V For 3.3V, 5.0V, and ADJ Versions V _{FEEDBACK} = 15V For 12V Versions	4.2		6	mA
I _{STBY}	Standby Quiescent Current	ON/OFF Pin = 0V	50		100 150	μA
I _{CL}	Current Limit		7	6.1 5.75	8.3 8.75	А
IL	Output Leakage Current	$V_{IN} = 40V$, ON/OFF Pin = 0V $V_{SWITCH} = 0V$	1		200 15	μA
		V _{SWITCH} = -1V	6			mA
R _{DS(ON)}	Switch On- Resistance	I _{SWITCH} = 5A	0.12		0.14 0.225	Ω
f _O	Oscillator Frequency	Measured at Switch Pin	260	225	280	kHz
D	Duty Cycle	Maximum Duty Cycle	91			%
		Minimum Duty Cycle	0			%
I _{BIAS}	Feedback Bias Current	V _{FEEDBACK} = 1.3V ADJ Version Only	85			nA
V _{ON/OFF}	ON/OFF Threshold Voltage		1.4	0.8	2.0	V
I _{ON/OFF}	ON/OFF Input Current	ON/OFF Input = 0V	20		45	μA
F _{SYNC}	Synchronization Frequency	V _{SYNC} (Pin 5)=3.5V, 50% Duty Cycle	400			KHz
V _{SYNC}	SYNC Threshold Voltage		1.4			V
θ_{JA}	Thermal	T Package, Junction to Ambient ⁽¹⁾	65			°C/W
θ_{JA}	Resistance	T Package, Junction to Ambient (2)	45			
θ_{JC}		T Package, Junction to Case	2			
θ_{JA}		S Package, Junction to Ambient ⁽³⁾	56			
θ_{JA}		S Package, Junction to Ambient ⁽⁴⁾	35			
θ_{JA}		S Package, Junction to Ambient ⁽⁵⁾	26			
θ_{JC}		S Package, Junction to Case	2			++
θ_{JA}		SD Package, Junction to Ambient ⁽⁶⁾	55			°C/W
θ_{JA}		SD Package, Junction to Ambient ⁽⁷⁾	29			3,77

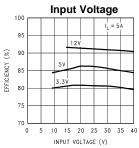
- Junction to ambient thermal resistance (no external heat sink) for the 7 lead TO-220 package mounted vertically, with ½ inch leads in a socket, or on a PC board with minimum copper area.
- (2) Junction to ambient thermal resistance (no external heat sink) for the 7 lead TO-220 package mounted vertically, with ½ inch leads soldered to a PC board containing approximately 4 square inches of (1 oz.) copper area surrounding the leads.
- (3) Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board area of 0.136 square inches (the same size as the TO-263 package) of 1 oz. (0.0014 in. thick) copper.
- (4) Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board area of 0.4896 square inches (3.6 times the area of the TO-263 package) of 1 oz. (0.0014 in. thick) copper.
- (5) Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board copper area of 1.0064 square inches (7.4 times the area of the TO-263 package) of 1 oz. (0.0014 in. thick) copper. Additional copper area will reduce thermal resistance further. See the thermal model in Switchers Made Simple software.
- (6) Junction to ambient thermal resistance for the 14-lead LLP mounted on a PC board copper area equal to the die attach paddle.
- (7) Junction to ambient thermal resistance for the 14-lead LLP mounted on a PC board copper area using 12 vias to a second layer of copper equal to die attach paddle. Additional copper area will reduce thermal resistance further. For layout recommendations, refer to Application Note AN-1187.

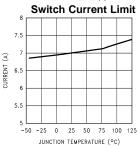


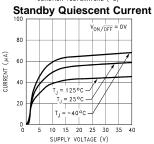
Typical Performance Characteristics



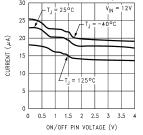
Efficiency vs

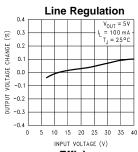




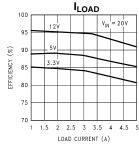


ON/OFF Pin Current (Sourcing)

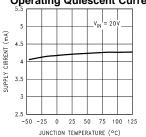




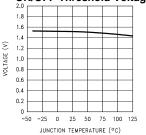
Efficiency vs



Operating Quiescent Current



ON/OFF Threshold Voltage



Switching Frequency

270

260

250

240

220

210

200

-50 -25 0 25 50 75 100 125

JUNCTION TEMPERATURE (°C)



Typical Performance Characteristics (continued)

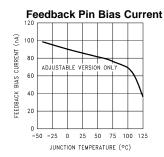
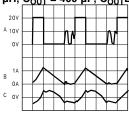


Figure .

Discontinuous Mode Switching Waveforms $V_{IN} = 20V,\, V_{OUT} = 5V,\, I_{LOAD} = 500~mA$ $L = 10~\mu H,\, \underline{C_{OUT}} = 400~\mu F,\, \underline{C_{OUT}} \text{ESR} = 13~m\Omega$

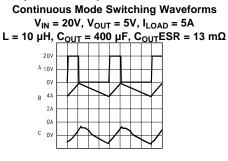


1 μsec/Div

A: V_{SW} Pin Voltage, 10 V/div. B: Inductor Current, 1 A/div

C: Output Ripple Voltage, 20 mV/div AC-Coupled

Figure 5. Horizontal Time Base: 1 µs//iv



A: V_{SW} Pin Voltage, 10 V/div.

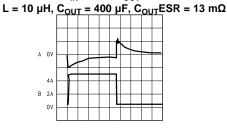
B: Inductor Current, 2 A/div

C: Output Ripple Voltage, 20 mV/div AC-Coupled

Figure 4. Horizontal Time Base: 1 µs/div

1 μsec/Div

Load Transient Response for Continuous Mode V_{IN} = 20V, V_{OUT} = 5V

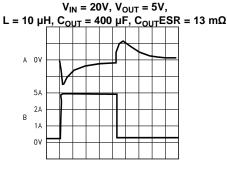


100 μsec/Div

A: Output Voltage, 100 mV//div, AC-Coupled. B: Load Current: 500 mA to 5A Load Pulse

Figure 6. Horizontal Time Base: 100 µs/div

Load Transient Response for Discontinuous Mode



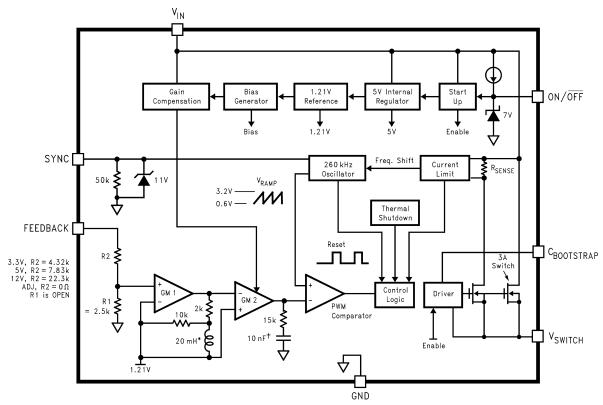
200 μsec/Div

A: Output Voltage, 100 mV/div, AC-Coupled. B: Load Current: 200 mA to 5A Load Pulse

Figure 7. Horizontal Time Base: 200 µs/div



Block Diagram



- * Active Inductor Patent Number 5,514,947
- † Active Capacitor Patent Number 5,382,918

Application Hints

The LM2677 provides all of the active functions required for a step-down (buck) switching regulator. The internal power switch is a DMOS power MOSFET to provide power supply designs with high current capability, up to 5A, and highly efficient operation.

The LM2677 is part of the SIMPLE SWITCHER family of power converters. A complete design uses a minimum number of external components, which have been pre-determined from a variety of manufacturers. Using either this data sheet or a design software program called **Switchers Made Simple** a complete switching power supply can be designed quickly. The software is provided free of charge and can be downloaded from National Semiconductor's Internet site located at http://www.national.com.

SWITCH OUTPUT

This is the output of a power MOSFET switch connected directly to the input voltage. The switch provides energy to an inductor, an output capacitor and the load circuitry under control of an internal pulse-width-modulator (PWM). The PWM controller is internally clocked by a fixed 260KHz oscillator. In a standard step-down application the duty cycle (Time ON/Time OFF) of the power switch is proportional to the ratio of the power supply output voltage to the input voltage. The voltage on pin 1 switches between Vin (switch ON) and below ground by the voltage drop of the external Schottky diode (switch OFF).

INPUT

The input voltage for the power supply is connected to pin 2. In addition to providing energy to the load the input voltage also provides bias for the internal circuitry of the LM2677. For guaranteed performance the input voltage must be in the range of 8V to 40V. For best performance of the power supply the input pin should always be bypassed with an input capacitor located close to pin 2.

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C BOOST

A capacitor must be connected from pin 3 to the switch output, pin 1. This capacitor boosts the gate drive to the internal MOSFET above Vin to fully turn it ON. This minimizes conduction losses in the power switch to maintain high efficiency. The recommended value for C Boost is 0.01µF.

GROUND

This is the ground reference connection for all components in the power supply. In fast-switching, high-current applications such as those implemented with the LM2677, it is recommended that a broad ground plane be used to minimize signal coupling throughout the circuit

SYNC

This input allows control of the switching clock frequency. If left open-circuited the regulator will be switched at the internal oscillator frequency, between 225KHz and 280KHz. An external clock can be used to force the switching frequency and thereby control the output ripple frequency of the regulator. This capability provides for consistent filtering of the output ripple from system to system as well as precise frequency spectrum positioning of the ripple frequency which is often desired in communications and radio applications. This external frequency must be greater than the LM2677 internal oscillator frequency, which could be as high as 280KHz, to prevent an erroneous reset of the internal ramp oscillator and PWM control of the power switch. The ramp oscillator is reset on the positive going edge of the sync input signal. It is recommended that the external TTL or CMOS compatible clock (between 0V and a level greater than 3V) be ac coupled to the sync input through a 100pf capacitor and a $1K\Omega$ resistor to ground at pin 5 as shown in Figure 8.

When the SYNC function is used, current limit frequency foldback is not active. Therefore, the device my not be fully protected against extreme output short circuit conditions. See Additional Application Information.

FEEDBACK

This is the input to a two-stage high gain amplifier, which drives the PWM controller. It is necessary to connect pin 6 to the actual output of the power supply to set the dc output voltage. For the fixed output devices (3.3V, 5V and 12V outputs), a direct wire connection to the output is all that is required as internal gain setting resistors are provided inside the LM2677. For the adjustable output version two external resistors are required to set the dc output voltage. For stable operation of the power supply it is important to prevent coupling of any inductor flux to the feedback input.

ON/OFF

This input provides an electrical ON/OFF control of the power supply. Connecting this pin to ground or to any voltage less than 0.8V will completely turn OFF the regulator. The current drain from the input supply when OFF is only 50µA. Pin 7 has an internal pull-up current source of approximately 20µA and a protection clamp zener diode of 7V to ground. When electrically driving the ON/OFF pin the high voltage level for the ON condition should not exceed the 6V absolute maximum limit. When ON/OFF control is not required pin 7 should be left open circuited.

DAP (LLP PACKAGE)

The Die Attach Pad (DAP) can and should be connected to PCB Ground plane/island. For CAD and assembly guidelines refer to Application Note AN-1187 at http://power.national.com.



DESIGN CONSIDERATIONS

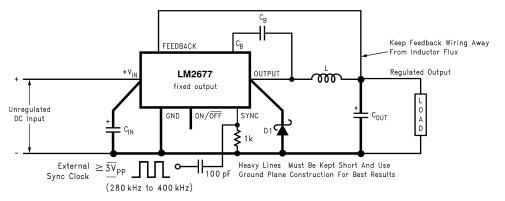


Figure 8. Basic circuit for fixed output voltage applications.

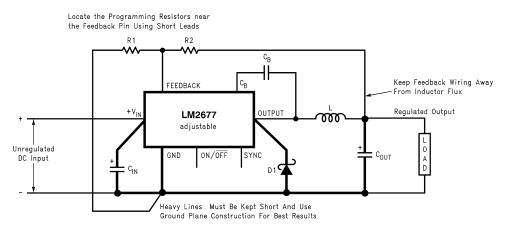


Figure 9. Basic circuit for adjustable output voltage applications

Power supply design using the LM2677 is greatly simplified by using recommended external components. A wide range of inductors, capacitors and Schottky diodes from several manufacturers have been evaluated for use in designs that cover the full range of capabilities (input voltage, output voltage and load current) of the LM2677. A simple design procedure using nomographs and component tables provided in this data sheet leads to a working design with very little effort. Alternatively, the design software, *LM267X Made Simple* (version 6.0), can also be used to provide instant component selection, circuit performance calculations for evaluation, a bill of materials component list and a circuit schematic.

The individual components from the various manufacturers called out for use are still just a small sample of the vast array of components available in the industry. While these components are recommended, they are not exclusively the only components for use in a design. After a close comparison of component specifications, equivalent devices from other manufacturers could be substituted for use in an application.

Important considerations for each external component and an explanation of how the nomographs and selection tables were developed follows.

INDUCTOR

The inductor is the key component in a switching regulator. For efficiency the inductor stores energy during the switch ON time and then transfers energy to the load while the switch is OFF.

Nomographs are used to select the inductance value required for a given set of operating conditions. The nomographs assume that the circuit is operating in continuous mode (the current flowing through the inductor never falls to zero). The magnitude of inductance is selected to maintain a maximum ripple current of 30% of the maximum load current. If the ripple current exceeds this 30% limit the next larger value is selected.

Product Folder Links: LM2677

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The inductors offered have been specifically manufactured to provide proper operation under all operating conditions of input and output voltage and load current. Several part types are offered for a given amount of inductance. Both surface mount and through-hole devices are available. The inductors from each of the three manufacturers have unique characteristics.

Renco: ferrite stick core inductors; benefits are typically lowest cost and can withstand ripple and transient peak currents above the rated value. These inductors have an external magnetic field, which may generate EMI.

Pulse Engineering: powdered iron toroid core inductors; these also can withstand higher than rated currents and, being toroid inductors, will have low EMI.

Coilcraft: ferrite drum core inductors; these are the smallest physical size inductors and are available only as surface mount components. These inductors also generate EMI but less than stick inductors.

OUTPUT CAPACITOR

The output capacitor acts to smooth the dc output voltage and also provides energy storage. Selection of an output capacitor, with an associated equivalent series resistance (ESR), impacts both the amount of output ripple voltage and stability of the control loop.

The output ripple voltage of the power supply is the product of the capacitor ESR and the inductor ripple current. The capacitor types recommended in the tables were selected for having low ESR ratings.

In addition, both surface mount tantalum capacitors and through-hole aluminum electrolytic capacitors are offered as solutions.

Impacting frequency stability of the overall control loop, the output capacitance, in conjunction with the inductor, creates a double pole inside the feedback loop. In addition the capacitance and the ESR value create a zero. These frequency response effects together with the internal frequency compensation circuitry of the LM2677 modify the gain and phase shift of the closed loop system.

As a general rule for stable switching regulator circuits it is desired to have the unity gain bandwidth of the circuit to be limited to no more than one-sixth of the controller switching frequency. With the fixed 260KHz switching frequency of the LM2677, the output capacitor is selected to provide a unity gain bandwidth of 40KHz maximum. Each recommended capacitor value has been chosen to achieve this result.

In some cases multiple capacitors are required either to reduce the ESR of the output capacitor, to minimize output ripple (a ripple voltage of 1% of Vout or less is the assumed performance condition), or to increase the output capacitance to reduce the closed loop unity gain bandwidth (to less than 40KHz). When parallel combinations of capacitors are required it has been assumed that each capacitor is the exact same part type.

The RMS current and working voltage (WV) ratings of the output capacitor are also important considerations. In a typical step-down switching regulator, the inductor ripple current (set to be no more than 30% of the maximum load current by the inductor selection) is the current that flows through the output capacitor. The capacitor RMS current rating must be greater than this ripple current. The voltage rating of the output capacitor should be greater than 1.3 times the maximum output voltage of the power supply. If operation of the system at elevated temperatures is required, the capacitor voltage rating may be de-rated to less than the nominal room temperature rating. Careful inspection of the manufacturer's specification for de-rating of working voltage with temperature is important.

INPUT CAPACITOR

Fast changing currents in high current switching regulators place a significant dynamic load on the unregulated power source. An input capacitor helps to provide additional current to the power supply as well as smooth out input voltage variations.

Like the output capacitor, the key specifications for the input capacitor are RMS current rating and working voltage. The RMS current flowing through the input capacitor is equal to one-half of the maximum dc load current so the capacitor should be rated to handle this. Paralleling multiple capacitors proportionally increases the current rating of the total capacitance. The voltage rating should also be selected to be 1.3 times the maximum input voltage. Depending on the unregulated input power source, under light load conditions the maximum input voltage could be significantly higher than normal operation and should be considered when selecting an input capacitor.



The input capacitor should be placed very close to the input pin of the LM2677. Due to relative high current operation with fast transient changes, the series inductance of input connecting wires or PCB traces can create ringing signals at the input terminal which could possibly propagate to the output or other parts of the circuitry. It may be necessary in some designs to add a small valued $(0.1\mu F$ to $0.47\mu F)$ ceramic type capacitor in parallel with the input capacitor to prevent or minimize any ringing.

CATCH DIODE

When the power switch in the LM2677 turns OFF, the current through the inductor continues to flow. The path for this current is through the diode connected between the switch output and ground. This forward biased diode clamps the switch output to a voltage less than ground. This negative voltage must be greater than -1V so a low voltage drop (particularly at high current levels) Schottky diode is recommended. Total efficiency of the entire power supply is significantly impacted by the power lost in the output catch diode. The average current through the catch diode is dependent on the switch duty cycle (D) and is equal to the load current times (1-D). Use of a diode rated for much higher current than is required by the actual application helps to minimize the voltage drop and power loss in the diode.

During the switch ON time the diode will be reversed biased by the input voltage. The reverse voltage rating of the diode should be at least 1.3 times greater than the maximum input voltage.

BOOST CAPACITOR

The boost capacitor creates a voltage used to overdrive the gate of the internal power MOSFET. This improves efficiency by minimizing the on resistance of the switch and associated power loss. For all applications it is recommended to use a $0.01\mu F/50V$ ceramic capacitor.

SYNC COMPONENTS

When synchronizing the LM2677 with an external clock it is recommended to connect the clock to pin 5 through a series 100pf capacitor and connect a $1K\Omega$ resistor to ground from pin 5. This RC network creates a short 100nS pulse on each positive edge of the clock to reset the internal ramp oscillator. The reset time of the oscillator is approximately 300nS.

ADDITIONAL APPLICATION INFORMATION

When the output voltage is greater than approximately 6V, and the duty cycle at minimum input voltage is greater than approximately 50%, the designer should exercise caution in selection of the output filter components. When an application designed to these specific operating conditions is subjected to a current limit fault condition, it may be possible to observe a large hysteresis in the current limit. This can affect the output voltage of the device until the load current is reduced sufficiently to allow the current limit protection circuit to reset itself.

Under current limiting conditions, the LM267x is designed to respond in the following manner:

- 1. At the moment when the inductor current reaches the current limit threshold, the ON-pulse is immediately terminated. This happens for any application condition.
- 2. However, the current limit block is also designed to momentarily reduce the duty cycle to below 50% to avoid subharmonic oscillations, which could cause the inductor to saturate.
- 3. Thereafter, once the inductor current falls below the current limit threshold, there is a small relaxation time during which the duty cycle progressively rises back above 50% to the value required to achieve regulation.

If the output capacitance is sufficiently 'large', it may be possible that as the output tries to recover, the output capacitor charging current is large enough to repeatedly re-trigger the current limit circuit before the output has fully settled. This condition is exacerbated with higher output voltage settings because the energy requirement of the output capacitor varies as the square of the output voltage (½CV²), thus requiring an increased charging current.

A simple test to determine if this condition might exist for a suspect application is to apply a short circuit across the output of the converter, and then remove the shorted output condition. In an application with properly selected external components, the output will recover smoothly.

Practical values of external components that have been experimentally found to work well under these specific operating conditions are $C_{OUT} = 47\mu F$, $L = 22\mu H$. It should be noted that even with these components, for a device's current limit of I_{CLIM} , the maximum load current under which the possibility of the large current limit hysteresis can be minimized is $I_{CLIM}/2$. For example, if the input is 24V and the set output voltage is 18V, then for a desired maximum current of 1.5A, the current limit of the chosen switcher must be confirmed to be at least 3A.

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Under extreme over-current or short circuit conditions, the LM267X employs frequency foldback in addition to the current limit. If the cycle-by-cycle inductor current increases above the current limit threshold (due to short circuit or inductor saturation for example) the switching frequency will be automatically reduced to protect the IC. Frequency below 100 KHz is typical for an extreme short circuit condition.

SIMPLE DESIGN PROCEDURE

Using the nomographs and tables in this data sheet (or use the available design software at http://www.national.com) a complete step-down regulator can be designed in a few simple steps.

Step 1: Define the power supply operating conditions:

Required output voltage

Maximum DC input voltage

Maximum output load current

Step 2: Set the output voltage by selecting a fixed output LM2677 (3.3V, 5V or 12V applications) or determine the required feedback resistors for use with the adjustable LM2677–ADJ

Step 3: Determine the inductor required by using one of the four nomographs, Figure 10 through Figure 13. Table 1 provides a specific manufacturer and part number for the inductor.

Step 4: Using Table 3 (fixed output voltage) or Table 6 (adjustable output voltage), determine the output capacitance required for stable operation. Table 2 provides the specific capacitor type from the manufacturer of choice.

Step 5: Determine an input capacitor from Table 4 for fixed output voltage applications. Use Table 2 to find the specific capacitor type. For adjustable output circuits select a capacitor from Table 2 with a sufficient working voltage (WV) rating greater than Vin max, and an rms current rating greater than one-half the maximum load current (2 or more capacitors in parallel may be required).

Step 6: Select a diode from Table 5. The current rating of the diode must be greater than I load max and the Reverse Voltage rating must be greater than Vin max.

Step 7: Include a 0.01µF/50V capacitor for Cboost in the design.

FIXED OUTPUT VOLTAGE DESIGN EXAMPLE

A system logic power supply bus of 3.3V is to be generated from a wall adapter which provides an unregulated DC voltage of 13V to 16V. The maximum load current is 2.5A. Through-hole components are preferred.

Step 1: Operating conditions are:

Vout = 3.3V

Vin max = 16V

Iload max = 2.5A

Step 2: Select an LM2677T-3.3. The output voltage will have a tolerance of

±2% at room temperature and ±3% over the full operating temperature range.

Step 3: Use the nomograph for the 3.3V device ,Figure 10. The intersection of the 16V horizontal line (V_{in} max) and the 2.5A vertical line (I_{load} max) indicates that L33, a 22µH inductor, is required.

From Table 1, L33 in a through-hole component is available from Renco with part number RL-1283-22-43 or part number PE-53933 from Pulse Engineering.

Step 4: Use Table 3 to determine an output capacitor. With a 3.3V output and a 22µH inductor there are four through-hole output capacitor solutions with the number of same type capacitors to be paralleled and an identifying capacitor code given. Table 2 provides the actual capacitor characteristics. Any of the following choices will work in the circuit:

1 x 220µF/10V Sanyo OS-CON (code C5)

1 x 1000µF/35V Sanyo MV-GX (code C10)

1 x 2200µF/10V Nichicon PL (code C5)



1 x 1000µF/35V Panasonic HFQ (code C7)

Step 5: Use Table 4 to select an input capacitor. With 3.3V output and $22\mu H$ there are three through-hole solutions. These capacitors provide a sufficient voltage rating and an rms current rating greater than 1.25A (1/2 I_{load} max). Again using Table 2 for specific component characteristics the following choices are suitable:

1 x 1000µF/63V Sanyo MV-GX (code C14)

1 x 820µF/63V Nichicon PL (code C24)

1 x 560µF/50V Panasonic HFQ (code C13)

Step 6: From Table 5 a 3A Schottky diode must be selected. For through-hole components 20V rated diodes are sufficient and 2 part types are suitable:

1N5820

SR302

Step 7: A 0.01µF capacitor will be used for Cboost.

ADJUSTABLE OUTPUT DESIGN EXAMPLE

In this example it is desired to convert the voltage from a two battery automotive power supply (voltage range of 20V to 28V, typical in large truck applications) to the 14.8VDC alternator supply typically used to power electronic equipment from single battery 12V vehicle systems. The load current required is 2A maximum. It is also desired to implement the power supply with all surface mount components.

Step 1: Operating conditions are:

Vout = 14.8V

Vin max = 28V

Iload max = 2A

Step 2: Select an LM2677S-ADJ. To set the output voltage to 14.9V two resistors need to be chosen (R1 and R2 in Figure 9). For the adjustable device the output voltage is set by the following relationship:

$$V_{OUT} = V_{FB} \left(1 + \frac{R_2}{R_1} \right) \tag{1}$$

Where V_{FB} is the feedback voltage of typically 1.21V.

A recommended value to use for R1 is 1K. In this example then R2 is determined to be:

$$R_2 = R_1 \left(\frac{V_{OUT}}{V_{FB}} - 1 \right) = 1 \text{ k}\Omega \left(\frac{14.8 \text{ V}}{1.21 \text{ V}} - 1 \right)$$
 (2)

 $R2 = 11.23K\Omega$

The closest standard 1% tolerance value to use is $11.3K\Omega$

This will set the nominal output voltage to 14.88V which is within 0.5% of the target value.

Step 3: To use the nomograph for the adjustable device, Figure 13, requires a calculation of the inductor Volt•microsecond constant (E•T expressed in V•µS) from the following formula:

$$E \cdot T = (V_{IN(MAX)} - V_{OUT} - V_{SAT}) \cdot \frac{V_{OUT} + V_{D}}{V_{IN(MAX)} - V_{SAT} + V_{D}} \cdot \frac{1000}{260} (V \cdot \mu s)$$
(3)

where V_{SAT} is the voltage drop across the internal power switch which is $R_{ds(ON)}$ times I_{load} . In this example this would be typically 0.15Ω x 2A or 0.3V and V_D is the voltage drop across the forward bisased Schottky diode, typically 0.5V. The switching frequency of 260KHz is the nominal value to use to estimate the ON time of the switch during which energy is stored in the inductor.

For this example E•T is found to be:

$$E \cdot T = (28 - 14.8 - 0.3) \cdot \frac{14.8 + 0.5}{28 - 0.3 + 0.5} \cdot \frac{1000}{260} (V \cdot \mu s)$$
 (4)

$$E \cdot T = (12.9V) \cdot \frac{15.3}{28.2} \cdot 3.85 (V \cdot \mu s) = 26.9 (V \cdot \mu s)$$
 (5)



Using Figure 13, the intersection of $27V \cdot \mu S$ horizontally and the 2A vertical line (I_{load} max) indicates that L38, a 68 μH inductor, should be used.

From Table 1, L38 in a surface mount component is available from Pulse Engineering with part number PE-54038S.

Step 4: Use Table 6 to determine an output capacitor. With a 14.8V output the 12.5 to 15V row is used and with a 68µH inductor there are three surface mount output capacitor solutions. Table 2 provides the actual capacitor characteristics based on the C Code number. Any of the following choices can be used:

1 x 33µF/20V AVX TPS (code C6)

1 x 47µF/20V Sprague 594 (code C8)

1 x 47µF/20V Kemet T495 (code C8)

Important Note: When using the adjustable device in low voltage applications (less than 3V output), if the nomograph, Figure 6, selects an inductance of $22\mu H$ or less, Table 6 does not provide an output capacitor solution. With these conditions the number of output capacitors required for stable operation becomes impractical. It is recommended to use either a $33\mu H$ or $47\mu H$ inductor and the output capacitors from Table 6.

Step 5: An input capacitor for this example will require at least a 35V WV rating with an rms current rating of 1A (1/2 lout max). From Table 2 it can be seen that C12, a $33\mu F/35V$ capacitor from Sprague, has the required voltage/current rating of the surface mount components.

Step 6: From Table 5 a 3A Schottky diode must be selected. For surface mount diodes with a margin of safety on the voltage rating one of five diodes can be used:

SK34

30BQ040

30WQ04F

MBRS340

MBRD340

Step 7: A 0.01µF capacitor will be used for Cboost.

LLP PACKAGE DEVICES

The LM2677 is offered in the 14 lead LLP surface mount package to allow for a significantly decreased footprint with equivalent power dissipation compared to the TO-263. For details on mounting and soldering specifications, refer to Application Note AN-1187.

Inductor Selection Guides

For Continuous Mode Operation

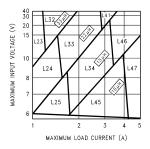


Figure 10. LM2677-3.3



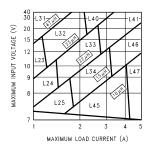


Figure 11. LM2677-5.0

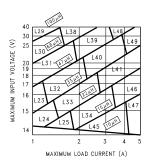


Figure 12. LM2677-12

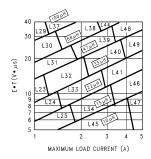


Figure 13. LM2677-ADJ

Table 1. Inductor Manufacturer Part Numbers (1)

Inductor	la di catana a	Cumant	Ren	со	Pulse E	ngineering	Coilcraft
Reference Number	Inductance (μΗ)	Current (A)	Through Hole	Surface Mount	Through Hole	Surface Mount	Surface Mount
L23	33	1.35	RL-5471-7	RL1500-33	PE-53823	PE-53823S	DO3316-333
L24	22	1.65	RL-1283-22-43	RL1500-22	PE-53824	PE-53824S	DO3316-223
L25	15	2.00	RL-1283-15-43	RL1500-15	PE-53825	PE-53825S	DO3316-153
L29	100	1.41	RL-5471-4	RL-6050-100	PE-53829	PE-53829S	DO5022P-104
L30	68	1.71	RL-5471-5	RL6050-68	PE-53830	PE-53830S	DO5022P-683
L31	47	2.06	RL-5471-6	RL6050-47	PE-53831	PE-53831S	DO5022P-473
L32	33	2.46	RL-5471-7	RL6050-33	PE-53932	PE-53932S	DO5022P-333
L33	22	3.02	RL-1283-22-43	RL6050-22	PE-53933	PE-53933S	DO5022P-223
L34	15	3.65	RL-1283-15-43	_	PE-53934	PE-53934S	DO5022P-153
L38	68	2.97	RL-5472-2	_	PE-54038	PE-54038S	_

(1) Assumes worst case maximum input voltage and load current for a given inductance value.



Table 1. Inductor Manufacturer Part Numbers⁽¹⁾ (continued)

Inductor	Industria	Current	Ren	со	Pulse E	ngineering	Coilcraft
Reference Number	Inductance (µH)	Thurstonia II ala		Surface Mount	Through Hole	Surface Mount	Surface Mount
L39	47	3.57	RL-5472-3	_	PE-54039	PE-54039S	_
L40	33	4.26	RL-1283-33-43	_	PE-54040	PE-54040S	_
L41	22	5.22	RL-1283-22-43	_	PE-54041	P0841	_
L44	68	3.45	RL-5473-3	_	PE-54044	_	_
L45	10	4.47	RL-1283-10-43	_	_	P0845	DO5022P-103HC
L46	15	5.60	RL-1283-15-43	_	_	P0846	DO5022P-153HC
L47	10	5.66	RL-1283-10-43	_	_	P0847	DO5022P-103HC
L48	47	5.61	RL-1282-47-43	_	_	P0848	_
L49	33	5.61	RL-1282-33-43	_		P0849	_

Table 2. Inductor Manufacturer Contact Numbers (1)

Coilcraft	Phone	(800) 322-2645
	FAX	(708) 639-1469
Coilcraft, Europe	Phone	+44 1236 730 595
	FAX	+44 1236 730 627
Pulse Engineering	Phone	(619) 674-8100
	FAX	(619) 674-8262
Pulse Engineering,	Phone	+353 93 24 107
Europe	FAX	+353 93 24 459
Renco Electronics	Phone	(800) 645-5828
	FAX	(516) 586-5562

⁽¹⁾ Assumes worst case maximum input voltage and load current for a given inductance value.

Capacitor Selection Guides (Assumes worst case maximum input voltage and load current for a given inductance value)

Table 3. Input and Output Capacitor Codes⁽¹⁾

_		Surface Mount											
Capacitor Reference Code	Δ	VX TPS Serie	es	Spra	Sprague 594D Series			Kemet T495 Series					
Reference oode	C (µF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)				
C1	330	6.3	1.15	120	6.3	1.1	100	6.3	0.82				
C2	100	10	1.1	220	6.3	1.4	220	6.3	1.1				
C3	220	10	1.15	68	10	1.05	330	6.3	1.1				
C4	47	16	0.89	150	10	1.35	100	10	1.1				
C5	100	16	1.15	47	16	1	150	10	1.1				
C6	33	20	0.77	100	16	1.3	220	10	1.1				
C7	68	20	0.94	180	16	1.95	33	20	0.78				
C8	22	25	0.77	47	20	1.15	47	20	0.94				
C9	10	35	0.63	33	25	1.05	68	20	0.94				
C10	22	35	0.66	68	25	1.6	10	35	0.63				
C11				15	35	0.75	22	35	0.63				
C12				33	35	1	4.7	50	0.66				
C13				15	50	0.9							

⁽¹⁾ Assumes worst case maximum input voltage and load current for a given inductance value.



Table 4. Input and Output Capacitor Codes (continued)⁽¹⁾

		Through Hole													
Capacitor Reference	Sanyo O	S-CON SA	Series	Sany	Sanyo MV-GX Series			Nichicon PL Series			Panasonic HFQ Series				
Code	C (μF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)			
C1	47	6.3	1	1000	6.3	8.0	680	10	0.8	82	35	0.4			
C2	150	6.3	1.95	270	16	0.6	820	10	0.98	120	35	0.44			
C3	330	6.3	2.45	470	16	0.75	1000	10	1.06	220	35	0.76			
C4	100	10	1.87	560	16	0.95	1200	10	1.28	330	35	1.01			
C5	220	10	2.36	820	16	1.25	2200	10	1.71	560	35	1.4			
C6	33	16	0.96	1000	16	1.3	3300	10	2.18	820	35	1.62			
C7	100	16	1.92	150	35	0.65	3900	10	2.36	1000	35	1.73			
C8	150	16	2.28	470	35	1.3	6800	10	2.68	2200	35	2.8			
C9	100	20	2.25	680	35	1.4	180	16	0.41	56	50	0.36			
C10	47	25	2.09	1000	35	1.7	270	16	0.55	100	50	0.5			
C11				220	63	0.76	470	16	0.77	220	50	0.92			
C12				470	63	1.2	680	16	1.02	470	50	1.44			
C13				680	63	1.5	820	16	1.22	560	50	1.68			
C14				1000	63	1.75	1800	16	1.88	1200	50	2.22			
C15							220	25	0.63	330	63	1.42			
C16							220	35	0.79	1500	63	2.51			
C17							560	35	1.43						
C18							2200	35	2.68						
C19							150	50	0.82						
C20							220	50	1.04						
C21							330	50	1.3						
C22							100	63	0.75						
C23							390	63	1.62						
C24							820	63	2.22						
C25							1200	63	2.51						

⁽¹⁾ Assumes worst case maximum input voltage and load current for a given inductance value.

Table 5. Capacitor Manufacturer Contact Numbers (1)

Nichicon	Phone	(847) 843-7500		
	FAX	(847) 843-2798		
Panasonic	Phone	(714) 373-7857		
	FAX	(714) 373-7102		
AVX	Phone	(845) 448-9411		
	FAX	(845) 448-1943		
Sprague/Vishay	Phone	(207) 324-4140		
	FAX	(207) 324-7223		
Sanyo	Phone	(619) 661-6322		
	FAX	(619) 661-1055		
Kemet	Phone	(864) 963-6300		
	FAX	(864) 963-6521		

⁽¹⁾ Assumes worst case maximum input voltage and load current for a given inductance value.



Table 6. Output Capacitors for Fixed Output Voltage Application⁽¹⁾

		Surface Mount								
Output Voltage (V)	Inductance (µH)	AVX TF	S Series	Sprague !	594D Series	Kemet T	495 Series			
(•)		No.	C Code	No.	C Code	No.	C Code			
	10	5	C1	5	C1	5	C2			
2.2	15	4	C1	4	C1	4	C3			
3.3	22	3	C2	2	C7	3	C4			
	33	1	C1	2	C7	3	C4			
	10	4	C2	4	C6	4	C4			
	15	3	C3	2	C7	3	C5			
5	22	3	C2	2	C7	3	C4			
	33	2	C2	2	C3	2	C4			
	47	2	C2	1	C7	2	C4			
	10	4	C5	3	C6	5	C9			
	15	3	C5	2	C7	4	C9			
	22	2	C5	2	C6	3	C8			
12	33	2	C5	1	C7	3	C8			
	47	2	C4	1	C6	2	C8			
	68	1	C5	1	C5	2	C7			
	100	1	C4	1	C5	1	C8			

(1) Assumes worst case maximum input voltage and load current for a given inductance value.

		Through Hole										
Output Voltage (V)	Inductance (µH)		S-CON SA eries	Sanyo M	V-GX Series	Nichicor	PL Series	Panasonio	HFQ Series			
		No.	C Code	No.	C Code	No.	C Code	No.	C Code			
	10	2	C5	2	C6	1	C8	2	C6			
2.2	15	2	C5	2	C5	1	C7	2	C5			
	22	1	C5	1	C10	1	C5	1	C7			
	33	1	C5	1	C10	1	C5	1	C7			
10	2	C4	2	C5	1	C6	2	C5				
	15	1	C5	1	C10	1	C5	1	C7			
5	22	1	C5	1	C9	1	C5	1	C5			
	33	1	C4	1	C5	1	C4	1	C4			
	47	1	C4	1	C4	1	C2	2	C4			
	10	2	C7	1	C10	1	C14	2	C4			
	15	1	C8	1	C6	1	C17	1	C5			
	22	1	C7	1	C5	1	C13	1	C5			
12	33	1	C7	1	C4	1	C12	1	C4			
	47	1	C7	1	C3	1	C11	1	C3			
	68	1	C6	1	C2	1	C10	1	C3			
	100	1	C6	1	C2	1	C9	1	C1			



Table 7. Input Capacitors for Fixed Output Voltage Application⁽¹⁾

				Surfac	ce Mount			
Output Voltage (V)	Inductance (µH)	AVX TF	PS Series	Sprague	594D Series	Kemet T495 Series		
(•)		No.	C Code	No.	C Code	No.	C Code	
	10	3	C7	2	C10	3	C9	
	15	*	*	3	C13	4	C12	
3.3	22	*	*	2	C13	3	C12	
	33	*	*	2	C13	3	C12	
	10	3	C4	2	C6	3	C9	
	15	4	C9	3	C12	4	C10	
5	22	*	*	3	C13	4	C12	
	33	*	*	2	C13	3	C12	
	47	*	*	1	C13	2	C12	
	10	4	C9	2	C10	4	C10	
	15	4	C8	2	C10	4	C10	
	22	4	C9	3	C12	4	C10	
12	33	*	*	3	C13	4	C12	
	47	*	*	2	C13	3	C12	
	68	*	*	2	C13	2	C12	
	100	*	*	1	C13	2	C12	

(1) Assumes worst case maximum input voltage and load current for a given inductance value.

			Through Hole										
Output Voltage (V)	Inductance (µH)		S-CON SA eries	Sanyo M	Sanyo MV-GX Series		n PL Series	Panasonio	Panasonic HFQ Series				
		No.	C Code	No.	C Code	No.	C Code	No.	C Code				
	10	2	C9	2	C8	1	C18	1	C8				
2.2	15	*	*	2	C13	1	C25	1	C16				
3.3	22	*	*	1	C14	1	C24	1	C16				
	33	*	*	1	C14	1	C24	1	C16				
	10	2	C7	2	C8	1	C25	1	C8				
	15	*	*	2	C8	1	C25	1	C8				
5	22	*	*	2	C13	1	C25	1	C16				
	33	*	*	1	C14	1	C23	1	C13				
	47	*	*	1	C12	1	C19	1	C11				
	10	2	C10	2	C8	1	C18	1	C8				
	15	2	C10	2	C8	1	C18	1	C8				
	22	*	*	2	C8	1	C18	1	C8				
12	33	*	*	2	C12	1	C24	1	C14				
	47	*	*	1	C14	1	C23	1	C13				
	68	*	*	1	C13	1	C21	1	C15				
	100	*	*	1	C11	1	C22	1	C11				

Table 8. Schottky Diode Selection Table

Reverse Voltage	Su	rface Mount	Through Hole			
(V)	3A	5A or More	3A	5A or More		
20V	SK32		1N5820			
			SR302			



Table 8. Schottky Diode Selection Table (continued)

Reverse Voltage	Sur	face Mount	Thre	ough Hole
(V)	3A	31DQ03 MBRB1545CT 11 6TQ045S MI 31 S MI 31DQ05	3A	5A or More
30V	SK33	MBRD835L	1N5821	
	30WQ03F		31DQ03	
40V	SK34	MBRB1545CT	1N5822	
	30BQ040	6TQ045S	MBR340	MBR745
	30WQ04F		31DQ04	80SQ045
	MBRS340		SR403	6TQ045
	MBRD340			
50V or More	SK35		MBR350	
	30WQ05F		31DQ05	
			SR305	

Table 9. Diode Manufacturer Contact Numbers

International Rectifier	Phone	(310) 322-3331
	FAX	(310) 322-3332
Motorola	Phone	(800) 521-6274
	FAX	(602) 244-6609
General Semiconductor	Phone	(516) 847-3000
	FAX	(516) 847-3236
Diodes, Inc.	Phone	(805) 446-4800
	FAX	(805) 446-4850

Table 10. Output Capacitors for Adjustable Output Voltage Applications

		Surface Mount								
Output Voltage (V)	Inductance (µH)	AVX T	PS Series	Sprague	594D Series	Kemet T	495 Series			
		No.	C Code	No.	C Code	No.	C Code			
1.21 to 2.50	33*	7	C1	6	C2	7	C3			
1.21 to 2.50	47*	5	C1	4	C2	5	C3			
2542275	33*	4	C1	3	C2	4	C3			
2.5 to 3.75	47*	3	C1	2	C2	3	C3			
	22	4	C1	3	C2	4	C3			
3.75 to 5	33	3	C1	2	C2	3	C3			
	47	2	C1	2	C2	2	C3			
	22	3	C2	1	C3	3	C4			
E 40 C 0E	33	2	C2	2	C3	2	C4			
5 to 6.25	47	2	C2	2	C3	2	C4			
	68	1	C2	1	C3	1	C4			
	22	3	C2	1	C4	3	C4			
C 05 to 7.5	33	2	C2	1	C3	2	C4			
6.25 to 7.5	47	1	C3	1	C4	1	C6			
	68	1	C2	1	C3	1	C4			
	33	2	C5	1	C6	2	C8			
7.5 to 40	47	1	C5	1	C6	2	C8			
7.5 to 10	68	1	C5	1	C6	1	C8			
	100	1	C4	1	C5	1	C8			

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Table 10. Output Capacitors for Adjustable Output Voltage Applications (continued)

		Surface Mount								
Output Voltage (V)	Inductance (µH)	AVX TI	PS Series	Sprague	594D Series	Kemet T	495 Series			
		No.	C Code	No.	C Code	No.	C Code			
	33	1	C5	1	C6	2	C8			
10 to 12.5	47	1	C5	1	C6	2	C8			
10 to 12.5	68	1	C5	1	C6	1	C8			
	100	1	C5	1	C6	1	C8			
	33	1	C6	1	C8	1	C8			
40 5 40 45	47	1	C6	1	C8	1	C8			
12.5 to 15	68	1	C6	1	C8	1	C8			
	100	1	C6	1	C8	1	C8			
	33	1	C8	1	C10	2	C10			
45 4- 20	47	1	C8	1	C9	2	C10			
15 to 20	68	1	C8	1	C9	2	C10			
	100	1	C8	1	C9	1	C10			
	33	2	C9	2	C11	2	C11			
00.4 00	47	1	C10	1	C12	1	C11			
20 to 30	68	1	C9	1	C12	1	C11			
	100	1	C9	1	C12	1	C11			
	10			4	C13	8	C12			
	15			3	C13	5	C12			
20.4- 27	22	No Value	es Available	2	C13	4	C12			
30 to 37	33			1	C13	3	C12			
	47			1	C13	2	C12			
	68			1	C13	2	C12			

Table 11. Output Capacitors for Adjustable Output Voltage Applications (continued)

		Through Hole									
Output Voltage (V)	Inductance (µH)		S-CON SA eries	Sanyo M	Sanyo MV-GX Series		PL Series	Panasonic HFQ Series			
		No.	C Code	No.	C Code	No.	C Code	No.	C Code		
4.04 += 0.50	33*	2	C3	5	C1	5	C3	3	С		
1.21 to 2.50	47*	2	C2	4	C1	3	C3	2	C5		
0.5.4- 0.75	33*	1	C3	3	C1	3	C1	2	C5		
2.5 to 3.75	47*	1	C2	2	C1	2	C3	1	C5		
	22	1	C3	3	C1	3	C1	2	C5		
3.75 to 5	33	1	C2	2	C1	2	C1	1	C5		
	47	1	C2	2	C1	1	C3	1	C5		
	22	1	C5	2	C6	2	C3	2	C5		
E 40 C 0E	33	1	C4	1	C6	2	C1	1	C5		
5 to 6.25	47	1	C4	1	C6	1	C3	1	C5		
	68	1	C4	1	C6	1	C1	1	C5		
	22	1	C5	1	C6	2	C1	1	C5		
6.25 to 7.5	33	1	C4	1	C6	1	C3	1	C5		
6.25 to 7.5	47	1	C4	1	C6	1	C1	1	C5		
	68	1	C4	1	C2	1	C1	1	C5		

Product Folder Links: *LM2677*

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Table 11. Output Capacitors for Adjustable Output Voltage Applications (continued) (continued)

		Through Hole									
Output Voltage (V)	Inductance (µH)		S-CON SA ries	Sanyo M	Sanyo MV-GX Series		PL Series	Panasonic HFQ Series			
		No.	C Code	No.	C Code	No.	C Code	No.	C Code		
	33	1	C7	1	C6	1	C14	1	C5		
7.5 4 - 40	47	1	C7	1	C6	1	C14	1	C5		
7.5 to 10	68	1	C7	1	C2	1	C14	1	C2		
	100	1	C7	1	C2	1	C14	1	C2		
	33	1	C7	1	C6	1	C14	1	C5		
40 4- 40 5	47	1	C7	1	C2	1	C14	1	C5		
10 to 12.5	68	1	C7	1	C2	1	C9	1	C2		
	100	1	C7	1	C2	1	C9	1	C2		
	33	1	C9	1	C10	1	C15	1	C2		
40.545.45	47	1	C9	1	C10	1	C15	1	C2		
12.5 to 15	68	1	C9	1	C10	1	C15	1	C2		
	100	1	C9	1	C10	1	C15	1	C2		
	33	1	C10	1	C7	1	C15	1	C2		
45.45.00	47	1	C10	1	C7	1	C15	1	C2		
15 to 20	68	1	C10	1	C7	1	C15	1	C2		
	100	1	C10	1	C7	1	C15	1	C2		
	33			1	C7	1	C16	1	C2		
00.4 - 00	47	No V	/alues	1	C7	1	C16	1	C2		
20 to 30	68	Ava	ilable	1	C7	1	C16	1	C2		
	100			1	C7	1	C16	1	C2		
	10			1	C12	1	C20	1	C10		
	15			1	C11	1	C20	1	C11		
20.44 27	22	No V	'alues	1	C11	1	C20	1	C10		
30 to 37	33	Ava	ilable	1	C11	1	C20	1	C10		
	47			1	C11	1	C20	1	C10		
	68			1	C11	1	C20	1	C10		





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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
LM2677S-12	ACTIVE	DDPAK	KTW	7	45	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -12	Samples
LM2677S-12/NOPB	ACTIVE	DDPAK	KTW	7	45	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -12	Samples
LM2677S-3.3	ACTIVE	DDPAK	KTW	7	45	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -3.3	Samples
LM2677S-3.3/NOPB	ACTIVE	DDPAK	KTW	7	45	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -3.3	Samples
LM2677S-5.0	ACTIVE	DDPAK	KTW	7	45	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -5.0	Samples
LM2677S-5.0/NOPB	ACTIVE	DDPAK	KTW	7	45	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -5.0	Samples
LM2677S-ADJ	ACTIVE	DDPAK	KTW	7	45	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -ADJ	Samples
LM2677S-ADJ/NOPB	ACTIVE	DDPAK	KTW	7	45	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -ADJ	Samples
LM2677SD-12	ACTIVE	VSON	NHM	14	250	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0002XB	Samples
LM2677SD-12/NOPB	ACTIVE	VSON	NHM	14	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	S0002XB	Samples
LM2677SD-3.3/NOPB	ACTIVE	VSON	NHM	14	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	S0002YB	Samples
LM2677SD-5.0	ACTIVE	VSON	NHM	14	250	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0002ZB	Samples
LM2677SD-5.0/NOPB	ACTIVE	VSON	NHM	14	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	S0002ZB	Samples
LM2677SD-ADJ	ACTIVE	VSON	NHM	14	250	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0003AB	Samples
LM2677SD-ADJ/NOPB	ACTIVE	VSON	NHM	14	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	S0003AB	Samples
LM2677SDX-12	ACTIVE	VSON	NHM	14	2500	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0002XB	Samples
LM2677SDX-12/NOPB	ACTIVE	VSON	NHM	14	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM			Samples
LM2677SDX-3.3	ACTIVE	VSON	NHM	14	2500	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0002YB	Samples





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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
LM2677SDX-5.0	ACTIVE	VSON	NHM	14	2500	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0002ZB	Samples
LM2677SDX-ADJ	ACTIVE	VSON	NHM	14	2500	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 125	S0003AB	Sample
LM2677SDX-ADJ/NOPB	ACTIVE	VSON	NHM	14	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 125	S0003AB	Sample
LM2677SX-12	ACTIVE	DDPAK	KTW	7	500	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -12	Sample
LM2677SX-12/NOPB	ACTIVE	DDPAK	KTW	7	500	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -12	Sample
LM2677SX-3.3	ACTIVE	DDPAK	KTW	7	500	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -3.3	Sample
LM2677SX-3.3/NOPB	ACTIVE	DDPAK	KTW	7	500	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -3.3	Sample
LM2677SX-5.0	ACTIVE	DDPAK	KTW	7	500	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -5.0	Sample
LM2677SX-5.0/NOPB	ACTIVE	DDPAK	KTW	7	500	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -5.0	Sample
LM2677SX-ADJ	ACTIVE	DDPAK	KTW	7	500	TBD	CU SNPB	Level-3-235C-168 HR	-40 to 125	LM2677S -ADJ	Sample
LM2677SX-ADJ/NOPB	ACTIVE	DDPAK	KTW	7	500	Pb-Free (RoHS Exempt)	CU SN	Level-3-245C-168 HR	-40 to 125	LM2677S -ADJ	Sample
LM2677T-12	ACTIVE	TO-220	NDZ	7	45	TBD	CU SNPB	Level-1-NA-UNLIM	-40 to 125	LM2677T -12	Sample
LM2677T-12/NOPB	ACTIVE	TO-220	NDZ	7	45	Green (RoHS & no Sb/Br)	CU SN	Level-1-NA-UNLIM	-40 to 125	LM2677T -12	Sample
LM2677T-3.3	ACTIVE	TO-220	NDZ	7	45	TBD	CU SNPB	Level-1-NA-UNLIM	-40 to 125	LM2677T -3.3	Sample
LM2677T-3.3/NOPB	ACTIVE	TO-220	NDZ	7	45	Green (RoHS & no Sb/Br)	CU SN	Level-1-NA-UNLIM	-40 to 125	LM2677T -3.3	Sample
LM2677T-5.0	ACTIVE	TO-220	NDZ	7	45	TBD	CU SNPB	Level-1-NA-UNLIM	-40 to 125	LM2677T -5.0	Sample
LM2677T-5.0/NOPB	ACTIVE	TO-220	NDZ	7	45	Green (RoHS & no Sb/Br)	CU SN	Level-1-NA-UNLIM	-40 to 125	LM2677T -5.0	Sample
LM2677T-ADJ	ACTIVE	TO-220	NDZ	7	45	TBD	CU SNPB	Level-1-NA-UNLIM	-40 to 125	LM2677T -ADJ	Sample



PACKAGE OPTION ADDENDUM

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Orderable Device	Status	Package Type	Package Drawing		Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Diawing			(2)		(3)		(4)	
LM2677T-ADJ/NOPB	ACTIVE	TO-220	NDZ	7	45	Green (RoHS	CU SN	Level-1-NA-UNLIM	-40 to 125	LM2677T	Samples
						& no Sb/Br)				-ADJ	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Only one of markings shown within the brackets will appear on the physical device.

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PACKAGE MATERIALS INFORMATION

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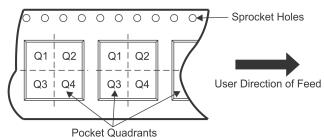
TAPE AND REEL INFORMATION





Α0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



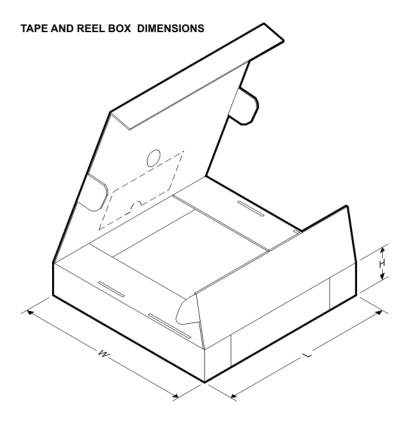
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2677SD-12	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SD-12/NOPB	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SD-3.3/NOPB	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SD-5.0	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SD-5.0/NOPB	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SD-ADJ	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SD-ADJ/NOPB	VSON	NHM	14	250	178.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SDX-12	VSON	NHM	14	2500	330.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SDX-12/NOPB	VSON	NHM	14	2500	330.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SDX-3.3	VSON	NHM	14	2500	330.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SDX-5.0	VSON	NHM	14	2500	330.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SDX-ADJ	VSON	NHM	14	2500	330.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SDX-ADJ/NOPB	VSON	NHM	14	2500	330.0	16.4	5.3	6.3	1.5	12.0	16.0	Q1
LM2677SX-12	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2677SX-12/NOPB	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2677SX-3.3	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2677SX-3.3/NOPB	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2677SX-5.0	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2

PACKAGE MATERIALS INFORMATION

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Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2677SX-5.0/NOPB	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2677SX-ADJ	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2677SX-ADJ/NOPB	DDPAK	KTW	7	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM2677SD-12	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SD-12/NOPB	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SD-3.3/NOPB	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SD-5.0	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SD-5.0/NOPB	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SD-ADJ	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SD-ADJ/NOPB	VSON	NHM	14	250	203.0	190.0	41.0
LM2677SDX-12	VSON	NHM	14	2500	349.0	337.0	45.0
LM2677SDX-12/NOPB	VSON	NHM	14	2500	349.0	337.0	45.0
LM2677SDX-3.3	VSON	NHM	14	2500	349.0	337.0	45.0
LM2677SDX-5.0	VSON	NHM	14	2500	349.0	337.0	45.0
LM2677SDX-ADJ	VSON	NHM	14	2500	349.0	337.0	45.0
LM2677SDX-ADJ/NOPB	VSON	NHM	14	2500	349.0	337.0	45.0
LM2677SX-12	DDPAK	KTW	7	500	358.0	343.0	63.0



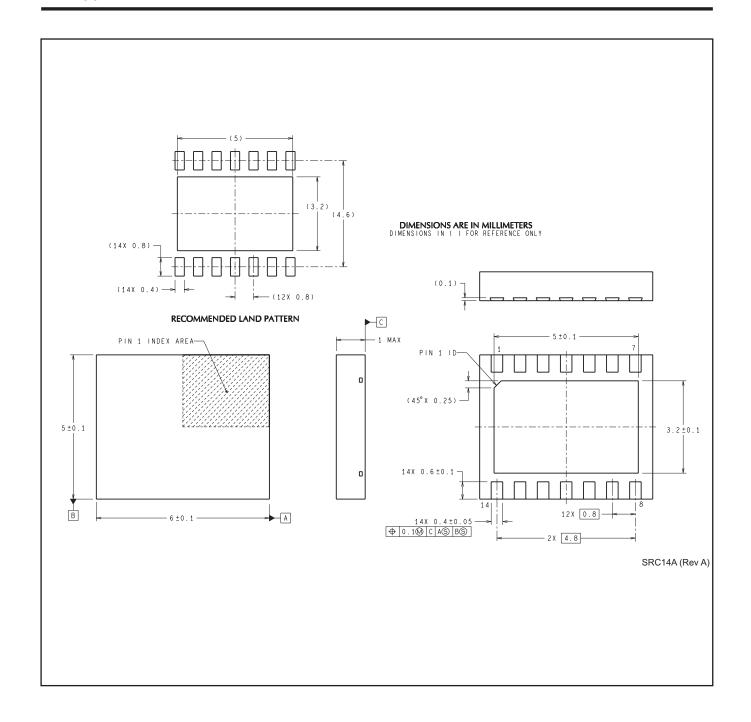
PACKAGE MATERIALS INFORMATION

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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM2677SX-12/NOPB	DDPAK	KTW	7	500	358.0	343.0	63.0
LM2677SX-3.3	DDPAK	KTW	7	500	358.0	343.0	63.0
LM2677SX-3.3/NOPB	DDPAK	KTW	7	500	358.0	343.0	63.0
LM2677SX-5.0	DDPAK	KTW	7	500	358.0	343.0	63.0
LM2677SX-5.0/NOPB	DDPAK	KTW	7	500	358.0	343.0	63.0
LM2677SX-ADJ	DDPAK	KTW	7	500	358.0	343.0	63.0
LM2677SX-ADJ/NOPB	DDPAK	KTW	7	500	358.0	343.0	63.0











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