

## LM2901EP Low Power Low Offset Voltage Quad Comparators

Check for Samples: [LM2901EP](#)

### FEATURES

- Wide supply voltage range
- LM2901: 2 to 36 V<sub>DC</sub> or  $\pm 1$  to  $\pm 18$  V<sub>DC</sub>
- Very low supply current drain (0.8 mA) — independent of supply voltage
- Low input biasing current: 25 nA
- Low input offset current:  $\pm 5$  nA
- Offset voltage:  $\pm 3$  mV
- Input common-mode voltage range includes GND
- Differential input voltage range equal to the power supply voltage
- Low output saturation voltage: 250 mV at 4 mA

- Output voltage compatible with TTL, DTL, ECL, MOS and CMOS logic systems

### ADVANTAGES

- High precision comparator
- Reduced V<sub>OS</sub> drift over temperature
- Eliminates need for dual supplies
- Allows sensing near GND
- Compatible with all forms of logic
- Power drain suitable for battery operation

### DESCRIPTION

The LM2901EP consists of four independent precision voltage comparators with an offset voltage specification as low as 2 mV max for all four comparators. These were designed specifically to operate from a single power supply over a wide range of voltages. Operation from split power supplies is also possible and the low power supply current drain is independent of the magnitude of the power supply voltage. This comparator also has a unique characteristic in that the input common-mode voltage range includes ground, even though operated from a single power supply voltage.

Application areas include limit comparators, simple analog to digital converters; pulse, squarewave and time delay generators; wide range VCO; MOS clock timers; multivibrators and high voltage digital logic gates. The LM2901EP was designed to directly interface with TTL and CMOS. When operated from both plus and minus power supplies, it will directly interface with MOS logic— where the low power drain of the LM2901EP is a distinct advantage over standard comparators.

### ENHANCED PLASTIC

- Extended Temperature Performance of  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$
- Baseline Control - Single Fab & Assembly Site
- Process Change Notification (PCN)
- Qualification & Reliability Data
- Solder (PbSn) Lead Finish is standard
- Enhanced Diminishing Manufacturing Sources (DMS) Support

### Applications

- Selected Military Applications
- Selected Avionics Applications



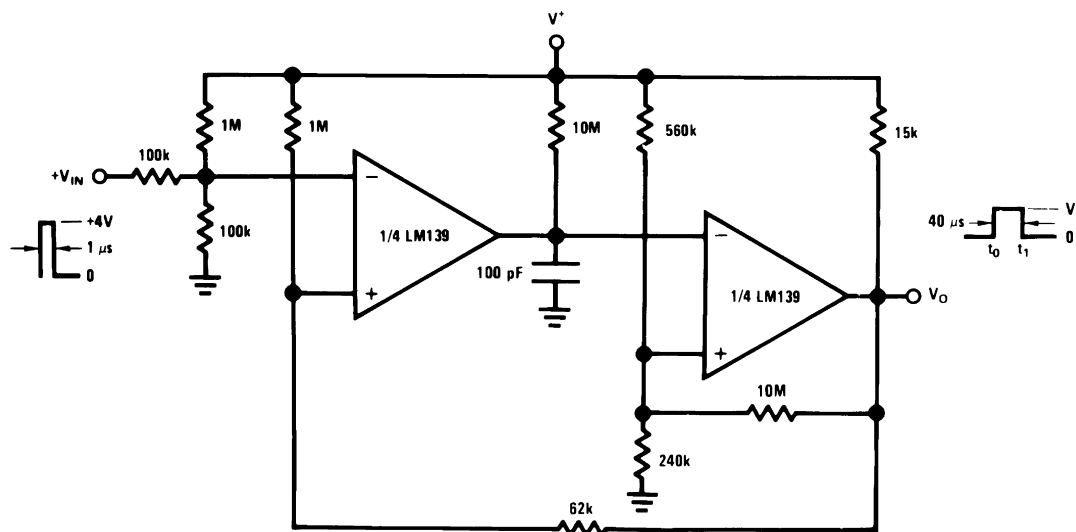
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## One-Shot Multivibrator with Input Lock Out



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## Absolute Maximum Ratings <sup>(1)</sup>

|   |   |
|---|---|
| Supply Voltage, $V^+$                                       | $36 V_{DC}$ or $\pm 18 V_{DC}$                  |
| Differential Input Voltage <sup>(2)</sup>                   | $36 V_{DC}$                                     |
| Input Voltage   | $-0.3 V_{DC}$ to $+36 V_{DC}$                   |
| Input Current ( $V_{IN} < -0.3 V_{DC}$ ),<br><sup>(3)</sup> | 50 mA   |
| Power Dissipation <sup>(4)</sup>                            |   |
| Molded DIP  | 1050 mW   |
| Small Outline Package                                       | 760 mW  |
| Output Short-Circuit to GND,<br><sup>(5)</sup>              | Continuous                                      |
| Storage Temperature Range                                   | $-65^{\circ}\text{C}$ to $+150^{\circ}\text{C}$ |
| Lead Temperature<br>(Soldering, 10 seconds)                 | $260^{\circ}\text{C}$                           |
| Operating Temperature Range                                 |   |
| LM2901  | $-40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$  |
| Soldering Information                                       |   |
| Dual-In-Line Package  |   |
| Soldering (10 seconds)                                      | $260^{\circ}\text{C}$                           |
| Small Outline Package                                       |   |
| Vapor Phase (60 seconds)                                    | $215^{\circ}\text{C}$                           |
| Infrared (15 seconds)                                       | $220^{\circ}\text{C}$                           |
| ESD rating (1.5 k $\Omega$ in series with 100 pF)           | 600V  |

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
- (2) Positive excursions of input voltage may exceed the power supply level. As long as the other voltage remains within the common-mode range, the comparator will provide a proper output state. The low input voltage state must not be less than  $-0.3 V_{DC}$  (or  $0.3 V_{DC}$  below the magnitude of the negative power supply, if used) (at  $25^{\circ}\text{C}$ ).
- (3) This input current will only exist when the voltage at any of the input leads is driven negative. It is due to the collector-base junction of the input PNP transistors becoming forward biased and thereby acting as input diode clamps. In addition to this diode action, there is also lateral NPN parasitic transistor action on the IC chip. This transistor action can cause the output voltages of the comparators to go to the  $V^+$  voltage level (or to ground for a large overdrive) for the time duration that an input is driven negative. This is not destructive and normal output states will re-establish when the input voltage, which was negative, again returns to a value greater than  $-0.3 V_{DC}$  (at  $25^{\circ}\text{C}$ ).
- (4) For operating at high temperatures, the LM2901EP must be derated based on a  $125^{\circ}\text{C}$  maximum junction temperature and a thermal resistance of  $95^{\circ}\text{C}/\text{W}$  which applies for the device soldered in a printed circuit board, operating in a still air ambient. The low bias dissipation and the "ON-OFF" characteristic of the outputs keeps the chip dissipation very small ( $P_D \leq 100 \text{ mW}$ ), provided the output transistors are allowed to saturate.
- (5) Short circuits from the output to  $V^+$  can cause excessive heating and eventual destruction. When considering short circuits to ground, the maximum output current is approximately 20 mA independent of the magnitude of  $V^+$ .

## Electrical Characteristics <sup>(1)</sup>

( $V^+ = 5 V_{DC}$ ,  $T_A = 25^\circ\text{C}$ , unless otherwise stated)

| Parameter                       | Conditions   | LM2901 |     |             | Units            |
|---------------------------------|--|--------|-----|-------------|------------------|
|                                 |  | Min    | Typ | Max         |                  |
| Input Offset Voltage            | <sup>(2)</sup>   |        | 2.0 | 7.0         | mV <sub>DC</sub> |
| Input Bias Current              | $I_{IN(+)}$ or $I_{IN(-)}$ with Output in Linear Range, <sup>(3)</sup> , $V_{CM}=0V$ |        | 25  | 250         | nA <sub>DC</sub> |
| Input Offset Current            | $I_{IN(+)} - I_{IN(-)}$ , $V_{CM} = 0V$  |        | 5   | 50          | nA <sub>DC</sub> |
| Input Common-Mode Voltage Range | $V^+ = 30 V_{DC}$  | 0      |     | $V^+ - 1.5$ | V <sub>DC</sub>  |
| Supply Current                  | $R_L = \infty$ on all Comparators,   |        | 0.8 | 2.0         | mA <sub>DC</sub> |
|                                 | $R_L = \infty$ , $V^+ = 36V$ ,   |        | 1.0 | 2.5         | mA <sub>DC</sub> |
| Voltage Gain                    | $R_L \geq 15 \text{ k}\Omega$ , $V^+ = 15 V_{DC}$                                    | 25     | 100 |             | V/mV             |
|                                 | $V_O = 1 V_{DC}$ to $11 V_{DC}$  |        |     |             |                  |
| Large Signal Response Time      | $V_{IN} = \text{TTL Logic Swing}$ , $V_{REF} =$                                      |        | 300 |             | ns               |
|                                 | $1.4 V_{DC}$ , $V_{RL} = 5 V_{DC}$ ,   |        |     |             |                  |
|                                 | $R_L = 5.1 \text{ k}\Omega$ ,  |        |     |             |                  |
| Response Time                   | $V_{RL} = 5 V_{DC}$ , $R_L = 5.1 \text{ k}\Omega$ ,                                  |        | 1.3 |             | $\mu\text{s}$    |
|                                 | <sup>(5)</sup>   |        |     |             |                  |
| Output Sink Current             | $V_{IN(-)} = 1 V_{DC}$ , $V_{IN(+)} = 0$ ,   | 6.0    | 16  |             | mA <sub>DC</sub> |
|                                 | $V_O \leq 1.5 V_{DC}$  |        |     |             |                  |
| Saturation Voltage              | $V_{IN(-)} = 1 V_{DC}$ , $V_{IN(+)} = 0$ ,   |        | 250 | 400         | mV <sub>DC</sub> |
|                                 | $I_{SINK} \leq 4 \text{ mA}$   |        |     |             |                  |
| Output Leakage Current          | $V_{IN(+)} = 1 V_{DC}$ , $V_{IN(-)} = 0$ ,   |        | 0.1 |             | nA <sub>DC</sub> |
|                                 | $V_O = 5 V_{DC}$   |        |     |             |                  |

- (1) "Testing and other quality control techniques are used to the extent deemed necessary to ensure product performance over the specified temperature range. Product may not necessarily be tested across the full temperature range and all parameters may not necessarily be tested. In the absence of specific PARAMETRIC testing, product performance is assured by characterization and/or design."
- (2) At output switch point,  $V_O \approx 1.4 V_{DC}$ ,  $R_S = 0\Omega$  with  $V^+$  from  $5 V_{DC}$  to  $30 V_{DC}$ ; and over the full input common-mode range ( $0 V_{DC}$  to  $V^+ - 1.5 V_{DC}$ ), at  $25^\circ\text{C}$ .
- (3) The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.
- (4) The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than  $0.3V$ . The upper end of the common-mode voltage range is  $V^+ - 1.5V$  at  $25^\circ\text{C}$ , but either or both inputs can go to  $+30 V_{DC}$  without damage independent of the magnitude of  $V^+$ .
- (5) The response time specified is a  $100 \text{ mV}$  input step with  $5 \text{ mV}$  overdrive. For larger overdrive signals  $300 \text{ ns}$  can be obtained, see typical performance characteristics section.

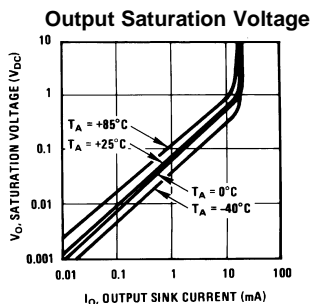
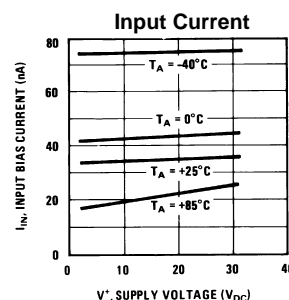
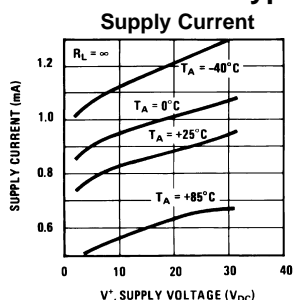
## Electrical Characteristics <sup>(1)</sup>

 $(V^+ = 5.0 V_{DC})$  <sup>(2)</sup>

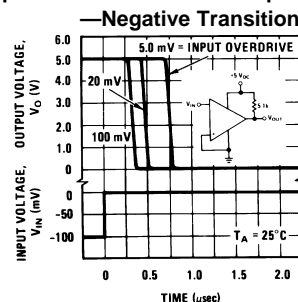
| Parameter                  | Conditions   | LM2901 |     |             | Units            |
|----------------------------|--|--------|-----|-------------|------------------|
|                            |  | Min    | Typ | Max         |                  |
| Input Offset Voltage       | <sup>(3)</sup>   |        | 9   | 15          | mV <sub>DC</sub> |
| Input Offset Current       | $I_{IN(+)} - I_{IN(-)}, V_{CM} = 0V$   |        | 50  | 200         | nA <sub>DC</sub> |
| Input Bias Current         | $I_{IN(+)}$ or $I_{IN(-)}$ with Output in Linear Range, $V_{CM} = 0V$ <sup>(4)</sup> |        | 200 | 500         | nA <sub>DC</sub> |
| Input Common-Mode          | $V^+ = 30 V_{DC}$  | 0      |     | $V^+ - 2.0$ | V <sub>DC</sub>  |
| Voltage Range              | <sup>(5)</sup>   |        |     |             |                  |
| Saturation Voltage         | $V_{IN(-)} = 1 V_{DC}, V_{IN(+)} = 0,$<br>$I_{SINK} \leq 4 \text{ mA}$               |        | 400 | 700         | mV <sub>DC</sub> |
| Output Leakage Current     | $V_{IN(+)} = 1 V_{DC}, V_{IN(-)} = 0,$<br>$V_O = 30 V_{DC}$                          |        |     | 1.0         | $\mu A_{DC}$     |
| Differential Input Voltage | Keep all $V_{IN}$ 's $\geq 0 V_{DC}$ (or $V^-$ , if used), <sup>(6)</sup>            |        |     | 36          | V <sub>DC</sub>  |

- (1) "Testing and other quality control techniques are used to the extent deemed necessary to ensure product performance over the specified temperature range. Product may not necessarily be tested across the full temperature range and all parameters may not necessarily be tested. In the absence of specific PARAMETRIC testing, product performance is assured by characterization and/or design."
- (2) These specifications are limited to  $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ , for the LM2901EP.
- (3) At output switch point,  $V_O \approx 1.4 V_{DC}$ ,  $R_S = 0\Omega$  with  $V^+$  from  $5 V_{DC}$  to  $30 V_{DC}$ ; and over the full input common-mode range ( $0 V_{DC}$  to  $V^+ - 1.5 V_{DC}$ ), at  $25^\circ\text{C}$ .
- (4) The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.
- (5) The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3V. The upper end of the common-mode voltage range is  $V^+ - 1.5V$  at  $25^\circ\text{C}$ , but either or both inputs can go to  $+30 V_{DC}$  without damage independent of the magnitude of  $V^+$ .
- (6) Positive excursions of input voltage may exceed the power supply level. As long as the other voltage remains within the common-mode range, the comparator will provide a proper output state. The low input voltage state must not be less than  $-0.3 V_{DC}$  (or  $0.3 V_{DC}$  below the magnitude of the negative power supply, if used) (at  $25^\circ\text{C}$ ).

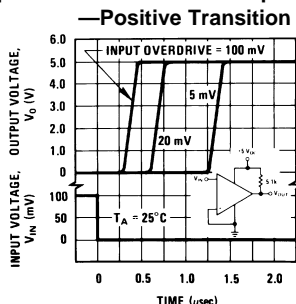
## Typical Performance Characteristics



### Response Time for Various Input Overdrives



### Response Time for Various Input Overdrives



## Application Hints

The LM2901EP is a high gain, wide bandwidth device which, like most comparators, can easily oscillate if the output lead is inadvertently allowed to capacitively couple to the inputs via stray capacitance. This shows up only during the output voltage transition intervals as the comparator changes states. Power supply bypassing is not required to solve this problem. Standard PC board layout is helpful as it reduces stray input-output coupling. Reducing this input resistors to  $< 10\text{ k}\Omega$  reduces the feedback signal levels and finally, adding even a small amount (1 to 10 mV) of positive feedback (hysteresis) causes such a rapid transition that oscillations due to stray feedback are not possible. Simply socketing the IC and attaching resistors to the pins will cause input-output oscillations during the small transition intervals unless hysteresis is used. If the input signal is a pulse waveform, with relatively fast rise and fall times, hysteresis is not required.

All pins of any unused comparators should be tied to the negative supply.

The bias network of the LM2901EP series establishes a drain current which is independent of the magnitude of the power supply voltage over the range of from  $2 V_{DC}$  to  $30 V_{DC}$ .

It is usually unnecessary to use a bypass capacitor across the power supply line.

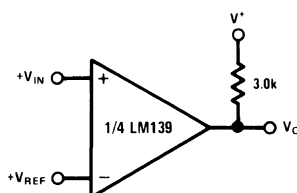
The differential input voltage may be larger than  $V^+$  without damaging the device. Protection should be provided to prevent the input voltages from going negative more than  $-0.3 V_{DC}$  (at  $25^\circ\text{C}$ ). An input clamp diode can be used as shown in the applications section.

The output of the LM2901EP is the uncommitted collector of a grounded-emitter NPN output transistor. Many collectors can be tied together to provide an output OR'ing function. An output pull-up resistor can be connected to any available power supply voltage within the permitted supply voltage range and there is no restriction on this voltage due to the magnitude of the voltage which is applied to the  $V^+$  terminal of the LM2901EP package. The output can also be used as a simple SPST switch to ground (when a pull-up resistor is not used). The amount of current which the output device can sink is limited by the drive available (which is independent of  $V^+$ ) and the  $\beta$  of this device. When the maximum current limit is reached (approximately 16 mA), the output transistor will come out of saturation and the output voltage will rise very rapidly. The output saturation voltage is limited by the approximately  $60\Omega R_{SAT}$  of the output transistor. The low offset voltage of the output transistor (1 mV) allows the output to clamp essentially to ground level for small load currents.

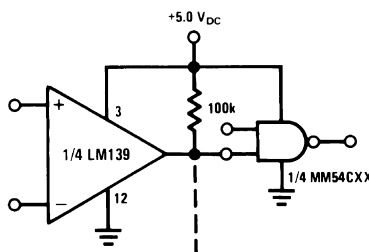
## Typical Applications

( $V^+ = 5.0 V_{DC}$ ) <sup>(1)</sup>

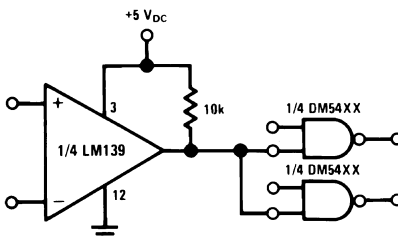
**Figure 1. Basic Comparator**



**Figure 2. Driving CMOS**



**Figure 3. Driving TTL**



(1) The LM139 within this data sheet's graphics is referenced because of it's a similarity to the LM2901, however is not offered in this data sheet.

$(V^+ = 5.0 V_{DC})^{(1)}$ 

Figure 4. AND Gate

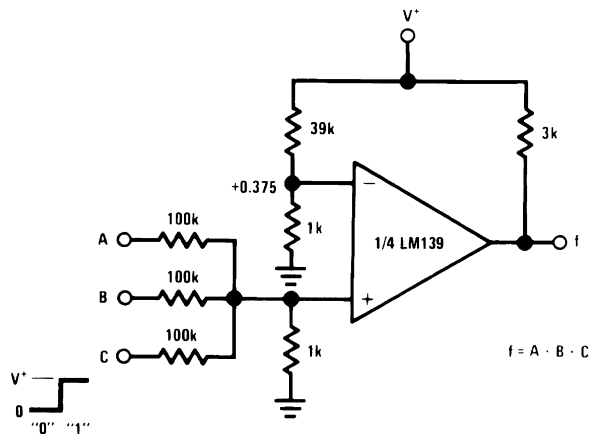
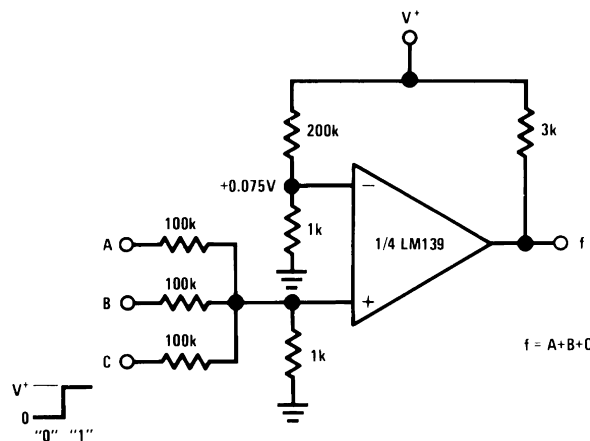


Figure 5. OR Gate



## Typical Applications

 $(V^+ = 15 V_{DC})^{(2)}$ 

(2) The LM139 within this data sheet's graphics is referenced because of it's a similarity to the LM2901, however is not offered in this data sheet.



( $V^+ = 15\text{ V}_{\text{DC}}$ ) <sup>(2)</sup>

Figure 6. One-Shot Multivibrator

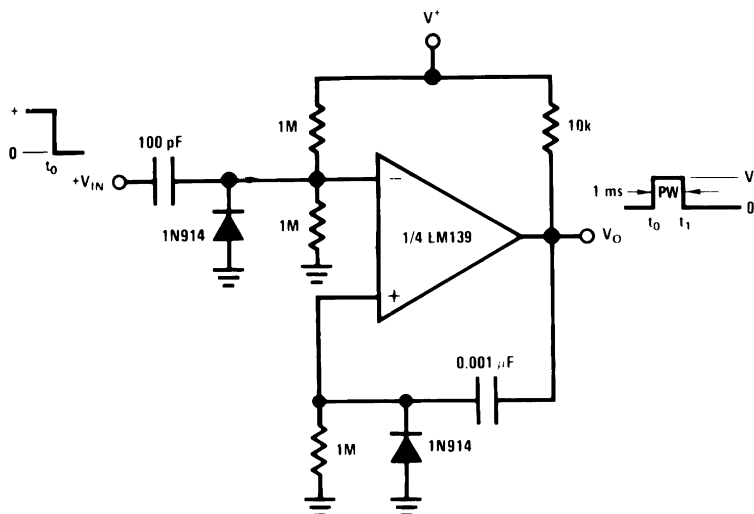


Figure 7. Bi-Stable Multivibrator

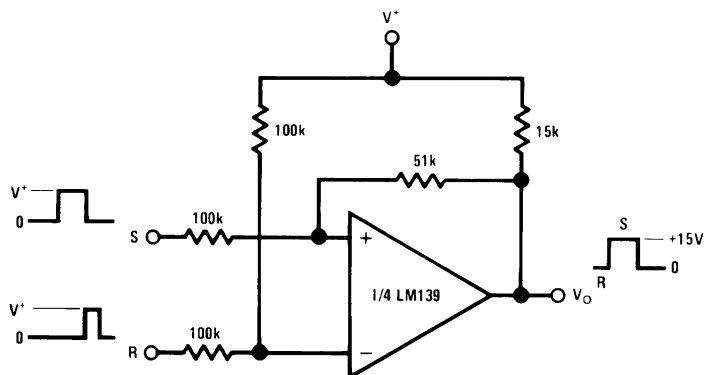
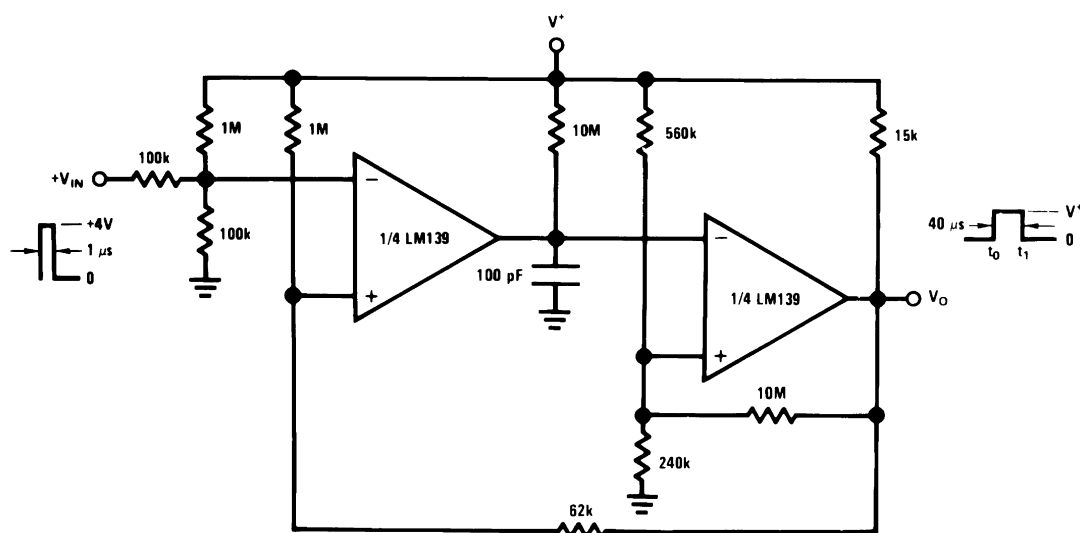


Figure 8. One-Shot Multivibrator with Input Lock Out



$(V^+ = 15\text{ V}_{\text{DC}})^{(2)}$ 

Figure 9. Pulse Generator

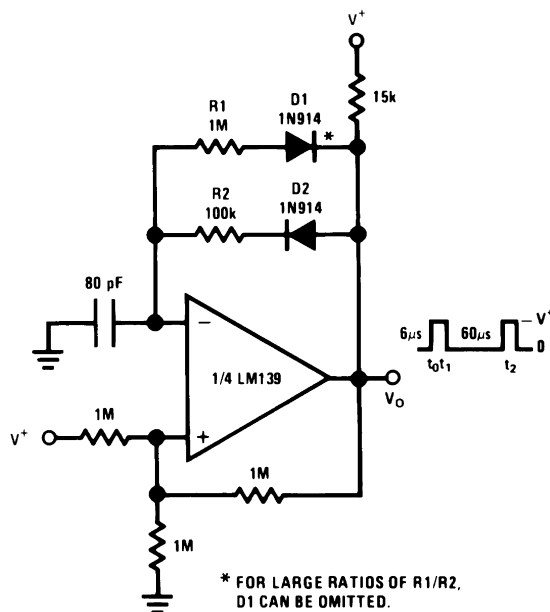
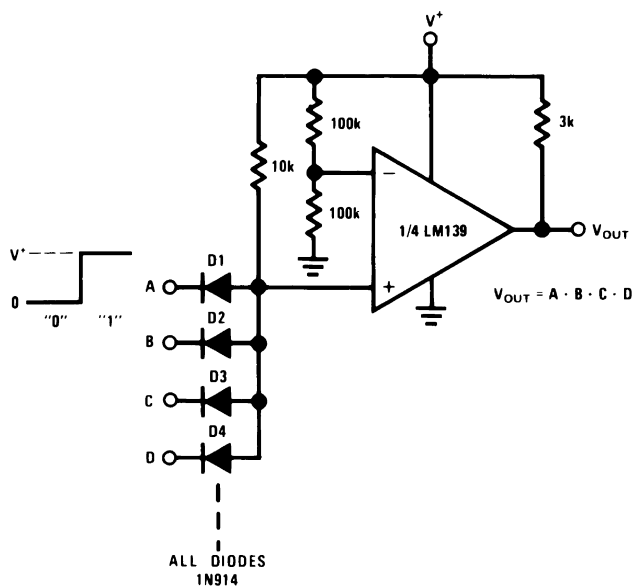
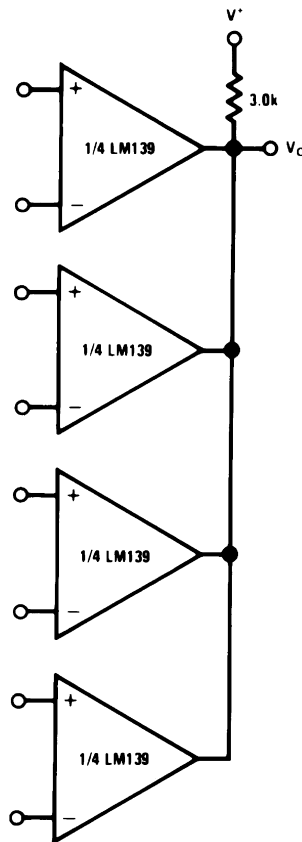


Figure 10. Large Fan-In AND Gate



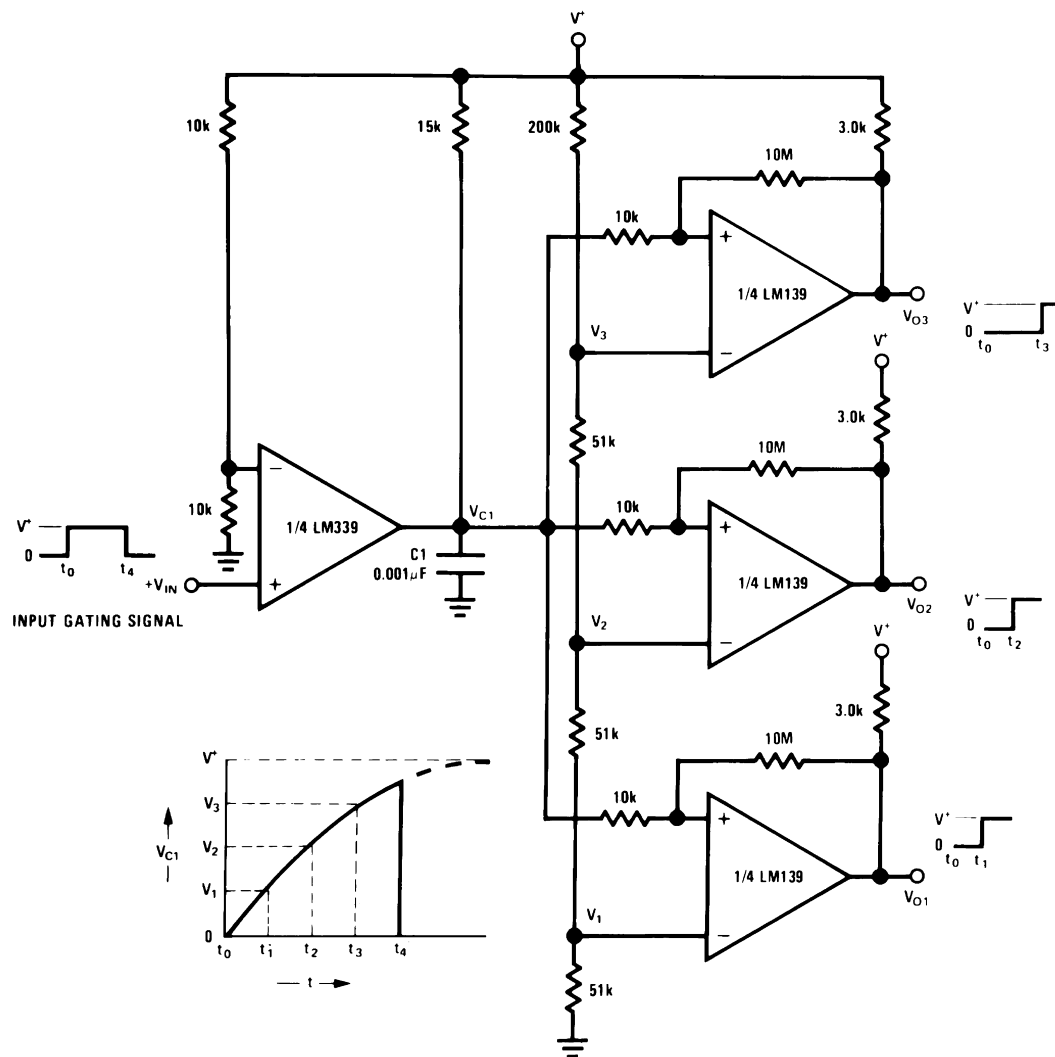
( $V^+ = 15\text{ V}_{\text{DC}}$ ) <sup>(2)</sup>

**Figure 11. ORing the Outputs**

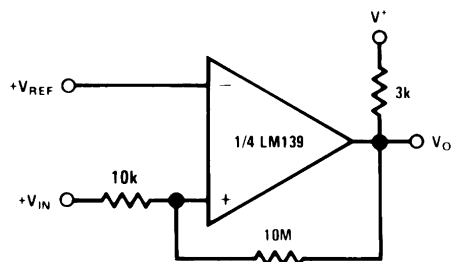


$$(V^+ = 15 V_{DC}) \quad (2)$$

### Figure 12. Time Delay Generator

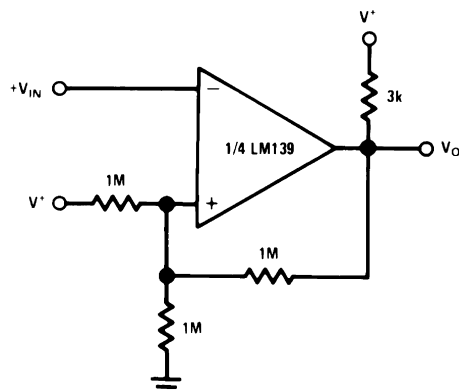


### Figure 13. Non-Inverting Comparator with Hysteresis

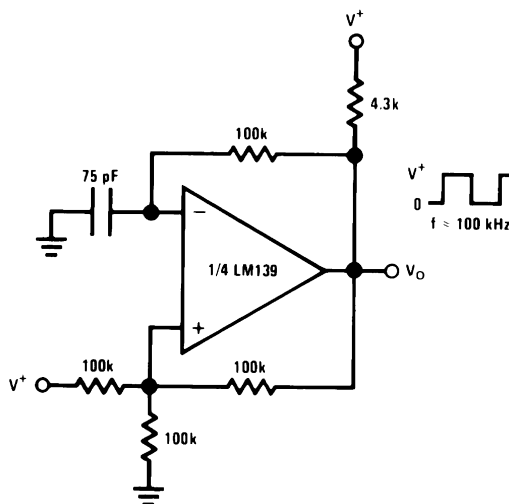


( $V^+ = 15\text{ V}_{\text{DC}}$ ) <sup>(2)</sup>

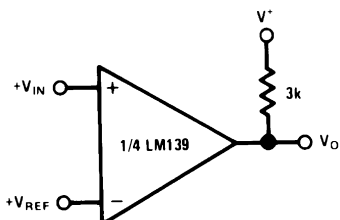
**Figure 14. Inverting Comparator with Hysteresis**

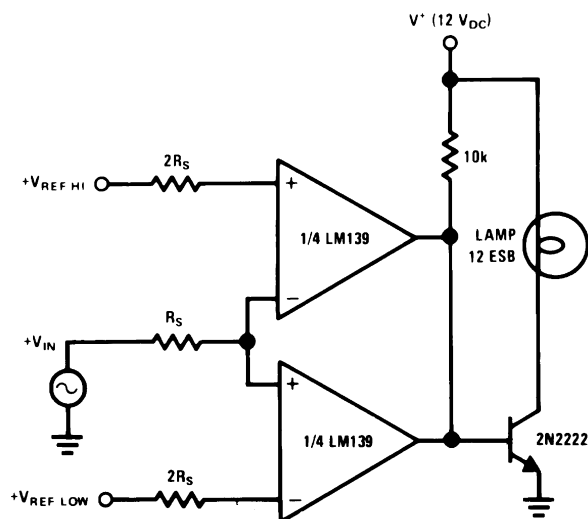
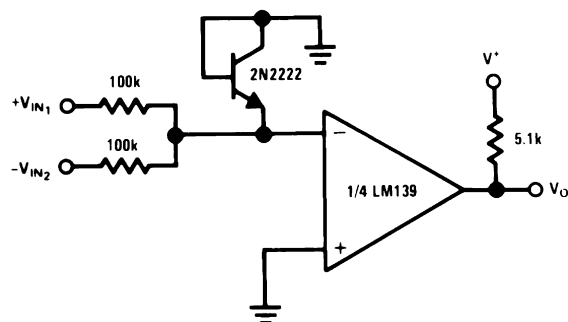
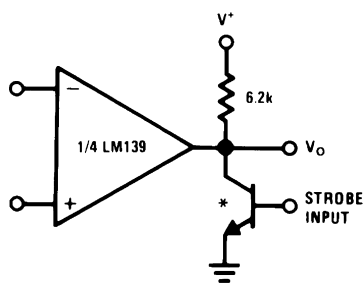


**Figure 15. Squarewave Oscillator**



**Figure 16. Basic Comparator**

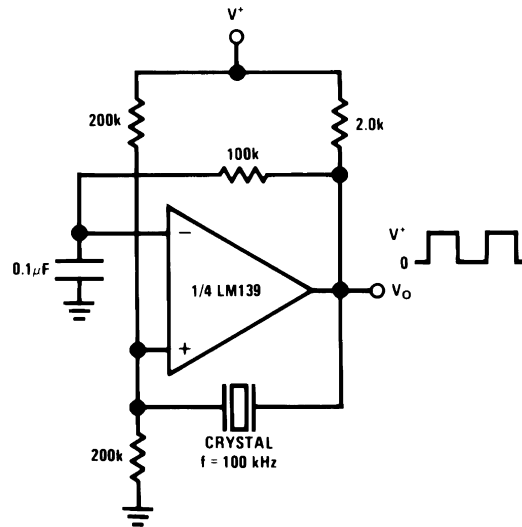


$(V^+ = 15\text{ V}_{\text{DC}})^{(2)}$ 
**Figure 17. Limit Comparator****Figure 18. Comparing Input Voltages of Opposite Polarity****Figure 19. Output Strobing**

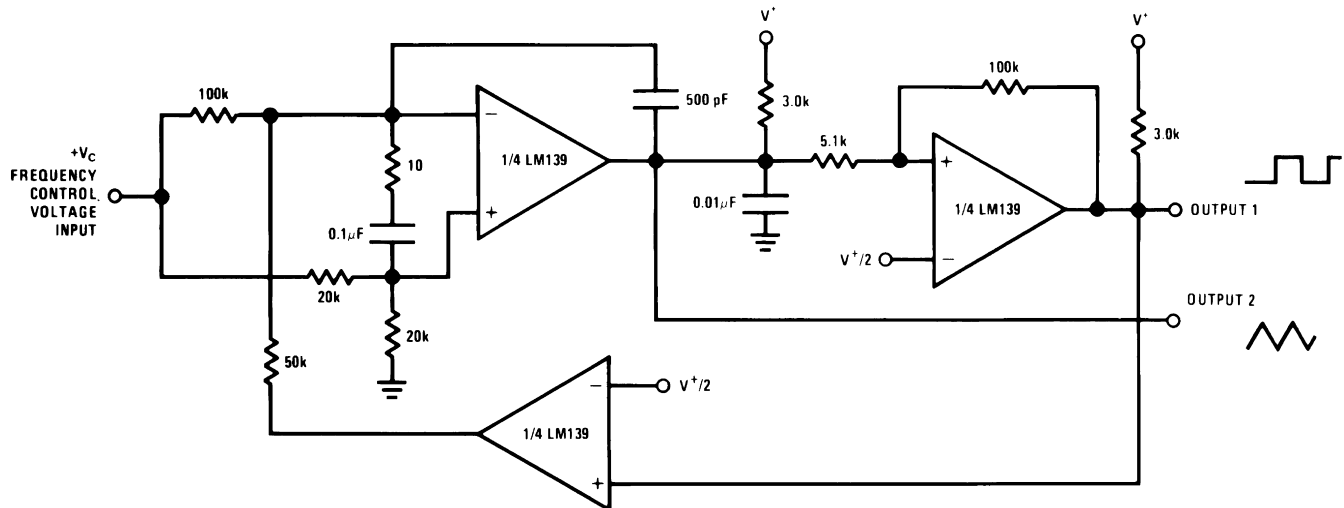
\* Or open-collector logic gate without pull-up resistor

( $V^+ = 15\text{ V}_{\text{DC}}$ ) <sup>(2)</sup>

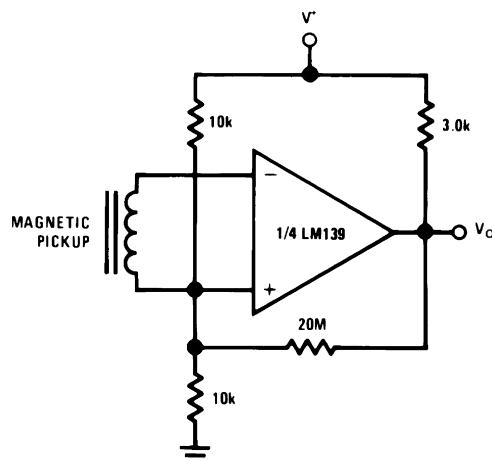
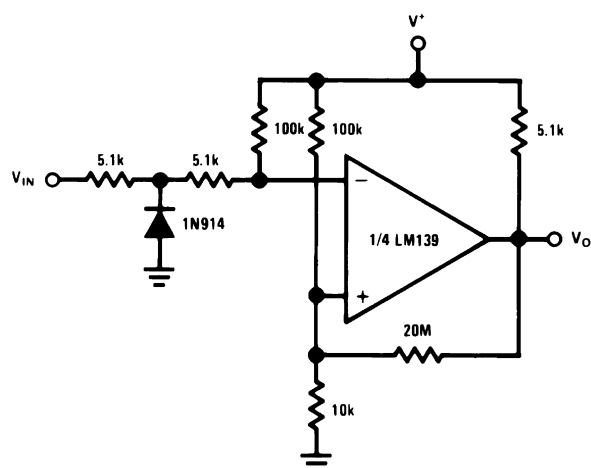
**Figure 20. Crystal Controlled Oscillator**



**Figure 21. Two-Decade High-Frequency VCO**



$V^+ = +30\text{ V}_{\text{DC}}$   
 $250\text{ mV}_{\text{DC}} \leq V_{\text{C}} \leq +50\text{ V}_{\text{DC}}$   
 $700\text{ Hz} \leq f_{\text{O}} \leq 100\text{ kHz}$

$(V^+ = 15\text{ V}_{\text{DC}})^{(2)}$ 
**Figure 22. Transducer Amplifier****Figure 23. Zero Crossing Detector (Single Power Supply)****Split-Supply Applications**
 $(V^+ = +15\text{ V}_{\text{DC}} \text{ and } V^- = -15\text{ V}_{\text{DC}})$



( $V^+ = +15\text{ V}_{\text{DC}}$  and  $V^- = -15\text{ V}_{\text{DC}}$ )

Figure 24. MOS Clock Driver

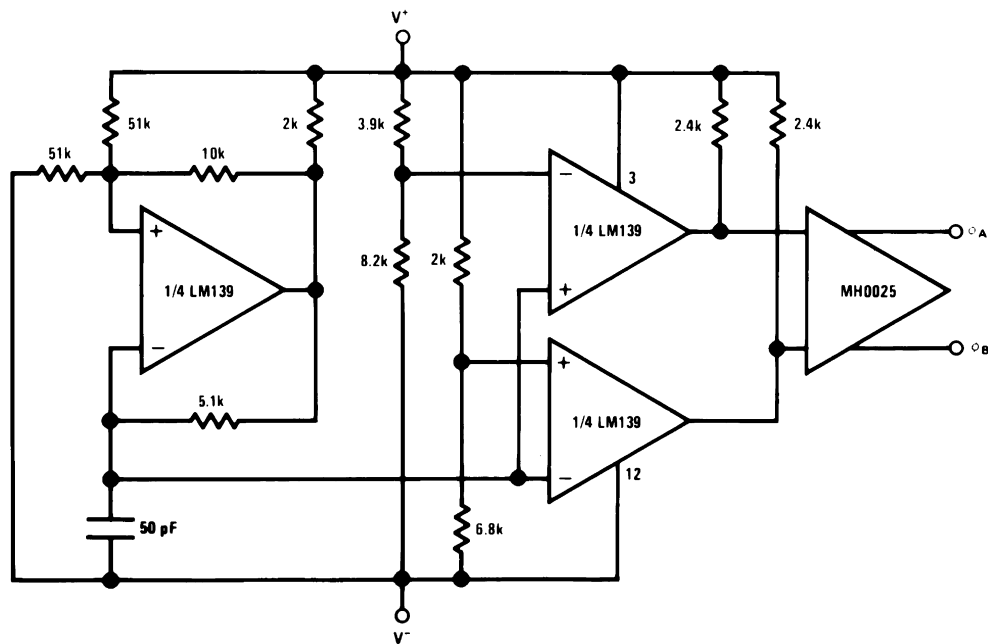


Figure 25. Zero Crossing Detector

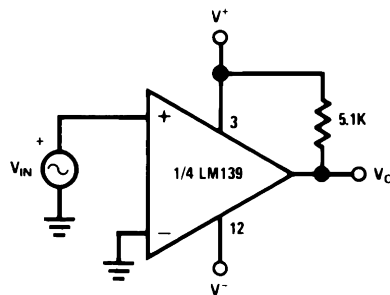
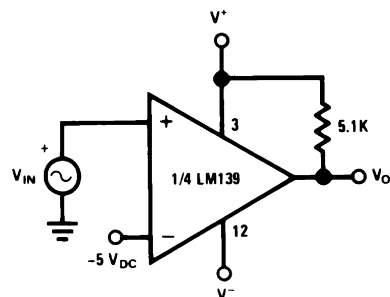
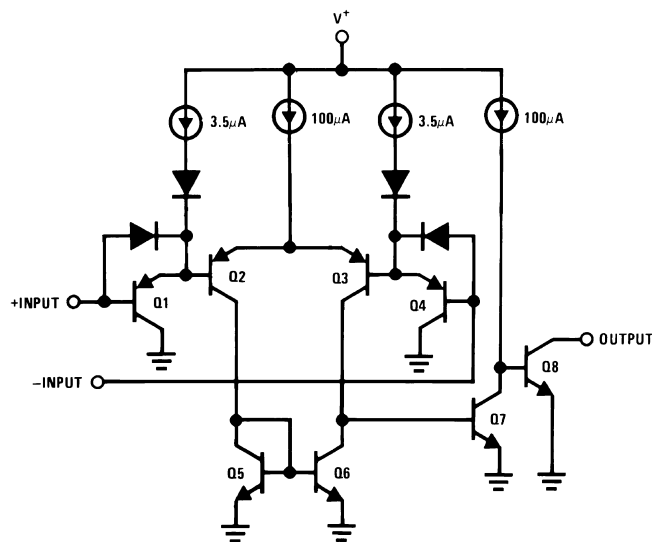


Figure 26. Comparator With a Negative Reference



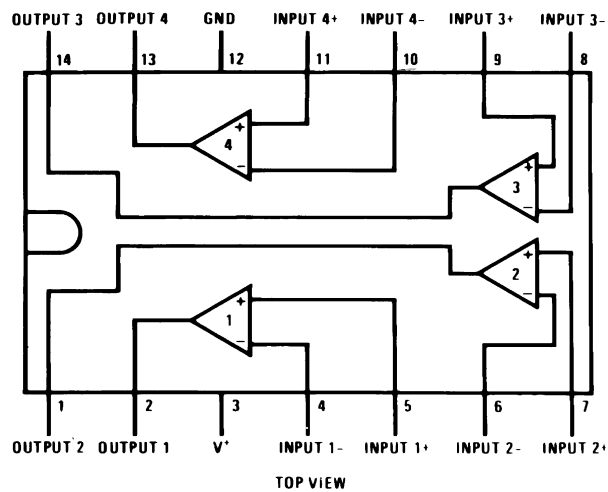
( $V^+ = +15\text{ V}_{\text{DC}}$  and  $V^- = -15\text{ V}_{\text{DC}}$ )

## Schematic Diagram



## Connection Diagrams

### Dual-In-Line Package



## IMPORTANT NOTICE

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